

OPERATING MANUAL
VIKING RANGER II
TRANSMITTER/EXCITER



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WASECA, MINNESOTA, U. S. A.

Johnson Viking Ranger II Transmitter Installation and Operation

The successful operation of any radio equipment is largely dependent upon the operator's understanding of the equipment. This operating instruction manual is set up in several parts, each with the purpose of making the operator more familiar with the Viking Ranger II. It is strongly recommended that this manual be read prior to attempting operation of the equipment. The main parts of the manual are:

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The Viking Ranger II should be given the good care usually accorded any other fine electronic instrument and in return will provide long trouble-free service. Periodic cleaning, dust removal, tube checking, etc. will maintain the equipment's appearance and performance.

WARNING

The voltages encountered in this piece of equipment are high enough to cause fatal injury! Practice safety rules until they are second nature. Always turn off the high voltage before making any adjustment inside the transmitter. Never depend on a bleeder resistor to discharge filter condensers. After the power is turned off, short circuit the high voltage circuit. Never operate the transmitter with any other than the recommended fuses in the primary circuit. The fuses will protect your equipment - in the case of accidental contact with the high voltage, they may save your life. If children have access to the open transmitter, always disable the primary circuit by removing the fuses, or the high voltage circuits by removing the rectifiers. Always remove the line cord plug from the power source when working inside the transmitter.

A. SET-UP INSTRUCTIONS FOR FACTORY WIRED TRANSMITTERS

1. After unpacking the transmitter, inspect thoroughly for any possible damage or mars from shipping. Claims against the carrier delivering the equipment must be made to the carrier agent at the point of delivery. **DO NOT SHIP DAMAGED EQUIPMENT BACK TO MANUFACTURER UNTIL NOTIFIED TO DO SO BY THE MANUFACTURERS. NOTIFY THE SERVICE DIVISION THAT A CLAIM IS BEING MADE AGAINST THE CARRIER.**
2. Remove the package containing plug P3 and antenna relay plug.
3. Installation of tubes - (In most cases tubes will already be in place.) Refer to Figure 1 for tube locations.
 - a. To facilitate removal of the VFO side cover, CH9, the crystal holder socket, J5, should be pivoted to one side to provide better access to the spade lug nearer the front panel. The mounting screw of J5, which can be seen in Figure 2, should be removed and the socket pivoted toward the right, in the direction of the bleeder resistor, R35. Be sure to replace the screw after the side cover has been secured in a later step.
 - (1) Place the OA2 tube, V2, in the socket nearest the front panel inside the VFO.
 - (2) Place the 6AU6 tube, V1, in the remaining socket inside the VFO. Do not replace side cover at this time. No tube shields are used in this transmitter.
 - b. Place V3, 6CL6 oscillator-multiplier, in socket XV3.
 - c. Place V4, 6CL6 Buffer-Multiplier, in socket XV4.
 - d. Place V12, 6AX5GT L.V. rectifier, in socket XV12.
 - e. Place V11, 5R4GY H.V. rectifier, in socket XV11.
 - f. Place V5, 6146 final amplifier, in socket XV5. Attach the plate cap E1.
 - g. Place V9 & V10, 7027 modulators, in sockets XV9 and XV10.
 - h. Place V8, 12AU7 audio driver, in socket X8.
 - i. Place V7, 12AX7 1st and 2nd audio speech amplifiers, in socket X7.
 - j. Place V6, 6AQ5 clamper tube, in socket X6.
 - k. Place V13, 12AU7 keyer tube, in socket X36.
 - l. Place V14, 6AL5 bias rectifier, in socket X35.
 - m. Place V15, 5763 6 Meter Doubler, in socket XV15.
4. Install knobs as follows (setscrews for all knobs are packaged separately and are installed at time of mounting):
 - a. Place a 1/4" I.D. deformed washer on all shafts except Buffer, Final and VFO shafts.
 - b. Install the large 2 3/8" knob, using one 10-32 setscrew, on the 1/4" shaft extending from the VFO planetary drive, being careful not to place the knob too close to the dial plate which would cause rubbing against the dial. Tighten the setscrew.
 - c. Install the 1 5/8" knob on the "BAND" switch shaft extension, using one 8-32 setscrew, making sure the knob marker coincides with the panel markings by turning the switch to the maximum counter-clockwise position and setting the knob marker to coincide with the 160 marker. Tighten the setscrew.
 - d. Install the seven single marker phenolic knobs as follows, using 8-32 setscrews:
 - (1) Turn the "DRIVE" control shaft fully counter-clockwise. Install one of the phenolic knobs with the marker at the "0" position and tighten the setscrew.
 - (2) Turn the "OPERATE" switch (SW4) to the counter-clockwise position, install one of the single marker phenolic knobs with the marker on the "OFF" position. Tighten the setscrew.
 - (3) Turn the "CRYSTAL-VFO" switch (SW2) to the counter-clockwise position, install one of the single marker phenolic knobs with the marker on the "XT1" position. Tighten the setscrew.

- A
 4.
 - d.
 - (4) Turn the "AUXILIARY COUPLING" switch (SW6) to the counter-clockwise position, install one of the single marker phenolic knobs, with the marker on numeral "1". Tighten the setscrew.
 - (5) Turn the "COUPLING" condenser (C9) into the fully meshed position. Install one of the single marker phenolic knobs with the marker on the "0" position. Tighten the setscrew.
 - (6) Turn the "AUDIO" gain control to the maximum counter-clockwise position, install one of the single marker phenolic knobs with the marker on the "0" position. Tighten the setscrew.
 - (7) Turn the "METER" switch (SW5) to the maximum counter-clockwise position. Install one of the single marker phenolic knobs, with marker on the "OFF" position. Tighten the setscrew.
 - e. Install the two 0 to 100 skirted knobs as follows, using 8-32 setscrews:
 - (1) Turn the BUFFER tuning condenser (C7) shaft until fully meshed. Install one of the knobs with the 0 directly under the maroon dot on the dial escutcheon. Tighten the setscrew.
 - (2) Turn the FINAL tuning condenser (C8) shaft until the condenser is fully meshed. Install one of the knobs with the 0 directly under the maroon dot on the dial escutcheon. Tighten the setscrew.
 - f. Before installing the crystal knob cover, it may be necessary to reform the contact fingers on the knob in order to reduce some of the pressure experienced in installing and removing the cover. This is done by pressing the contact fingers inward toward the center of the cover, a little at a time, until the cover can be installed relatively easy and still have a firm feeling when in place.
 - g. At this time, check the function of each knob to see if the indexing agrees with the markings on the panel (i.e. bandswitch on 160 when counter-clockwise and 0 when fully clockwise).
 5.
 - a. Check to see that both the 5 amp fuse and 3 amp Fusetron, or Slo-Blo fuse, are installed in the fused type line cord plug.
 - b. Place the "OPERATE" switch in the "OFF" position, and plug the line cord into a 117V AC receptacle. Plug P3 into the J6 socket on the back of transmitter.
 - c.
 - (1) Set the VFO pointer at mid-scale. Turn the drive control to position #5.
 - (2) Turn the "BAND" switch to 160.
 - (3) Turn the "CRYSTAL-VFO" switch to "VFO".
 - (4) Turn "OPERATE" switch to the "TUNE" position.
 - (5) Turn the meter switch to the "OSC" position.
 - a. Reading should be 24 to 32 ma. (This will be lower on 15 and 20 meters).
 - b. Turn the "CRYSTAL-VFO" switch to "XT1" position (with no crystal installed). Meter reading should be approximately 16 to 20 ma.
 - c. Install a crystal and check for increase in "OSC" current when switching from crystal to the blank crystal socket.
 - (6) Turn the "BAND" switch to "40" meter position and repeat steps A.5.c. (1 thru 5). Readings should be comparable.
 6. Turn the "BAND" switch back to the "160" position and turn the "OPERATE" switch to the "OFF" position. Remove the line cord from the 117V AC receptacle.
 7. At this time, refer to Section E of the Operating Manual, in order to determine if there has been an appreciable change in the calibration of the VFO due to possible changes in the VFO trimmer and padder condenser settings as a result of rough handling in shipment. It is doubtful that any pronounced deviation will occur when changing from one set of tubes to another in the VFO, since circuit design considerations guard against normal internal variances in the tubes. Occasional spot checks on both the 160 or 80

- A. 7. and the 40 meter bands against a frequency standard of reputable calibration accuracy will verify the VFO calibration accuracy. If the calibration is not accurate, proceed to re-calibrate as directed in Section E.
8. Read Sections B, Theory of Operation; C, Tuning Procedure, and D, Pi Network Tuning and Harmonic Suppression, in order to gain familiarity with the equipment before continuing tuneup.
9. Tune up the transmitter on all bands 160-6 meters, following the procedure in Section C, checking for grid drive, proper loading and operation on all bands
10. With the transmitter operating at normal full load (130 ma.) on "PHONE", check the "MOD" (modulator) plate current. It should read 55 to 70 ma. If the reading is not within this range, adjust the tap on resistor R35 - toward the panel for a lower current reading and away from the panel for a higher reading.
11. With the modulator current adjusted properly, the transmitter operating on "PHONE" and with microphone connected (left rear of cabinet, J1), advance the "AUDIO" gain control while speaking into the microphone. The "MOD" current should kick upward with voice peaks and the dummy light bulb should increase in brilliance. The current peak swings should not exceed 120-130 ma. for 100% modulation.
12. Operation of the clamper tube may be checked by leaving the transmitter on the phone or CW position, turning the VFO off by switching to an empty crystal socket position and watching the final plate current. The final plate current under excitation failure should be less than 50 ma.

B. JOHNSON VIKING RANGER II THEORY OF OPERATION

1. The Variable Frequency Oscillator is patterned after the famous Johnson Model 122, employing a series tuned Colpitts circuit. Two separate tank circuits are employed. One tank circuit covers the 1.75 mc to 2.0 mc frequency range and the other tank circuit covers the 7.000 mc to 7.425 mc range. The VFO tank circuits and output circuits are controlled by the internal VFO switch SW1. This switch is mechanically linked with the band switch insuring the proper VFO output frequency for the band selected by the band-switch. A high degree of stability is insured by proper circuit design, rigid construction, temperature compensation, and by voltage regulation in the VFO circuit. The plate circuit of the VFO is broad tuned to maintain a constant output level over the entire frequency excursion when employing the 40 meter tank. Additional circuit design considerations eliminate any interaction between the oscillator stage and succeeding stages in the RF exciter section.

On the 160 and 80 meter bands, the VFO output remains on the 160 meter tank. On the 40, 20, 15 and 10 meter bands the VFO output changes to the 40 meter tank. On the 6 meter band, additional capacitors, C4 and C91, are switched across the 40 meter tank, to lower the VFO frequency to a harmonic relation to the 6 meter band. In "VFO" operation both the VFO and the first frequency multiplier are keyed for CW transmission.

B. 2. Viking Ranger II Block Diagram.

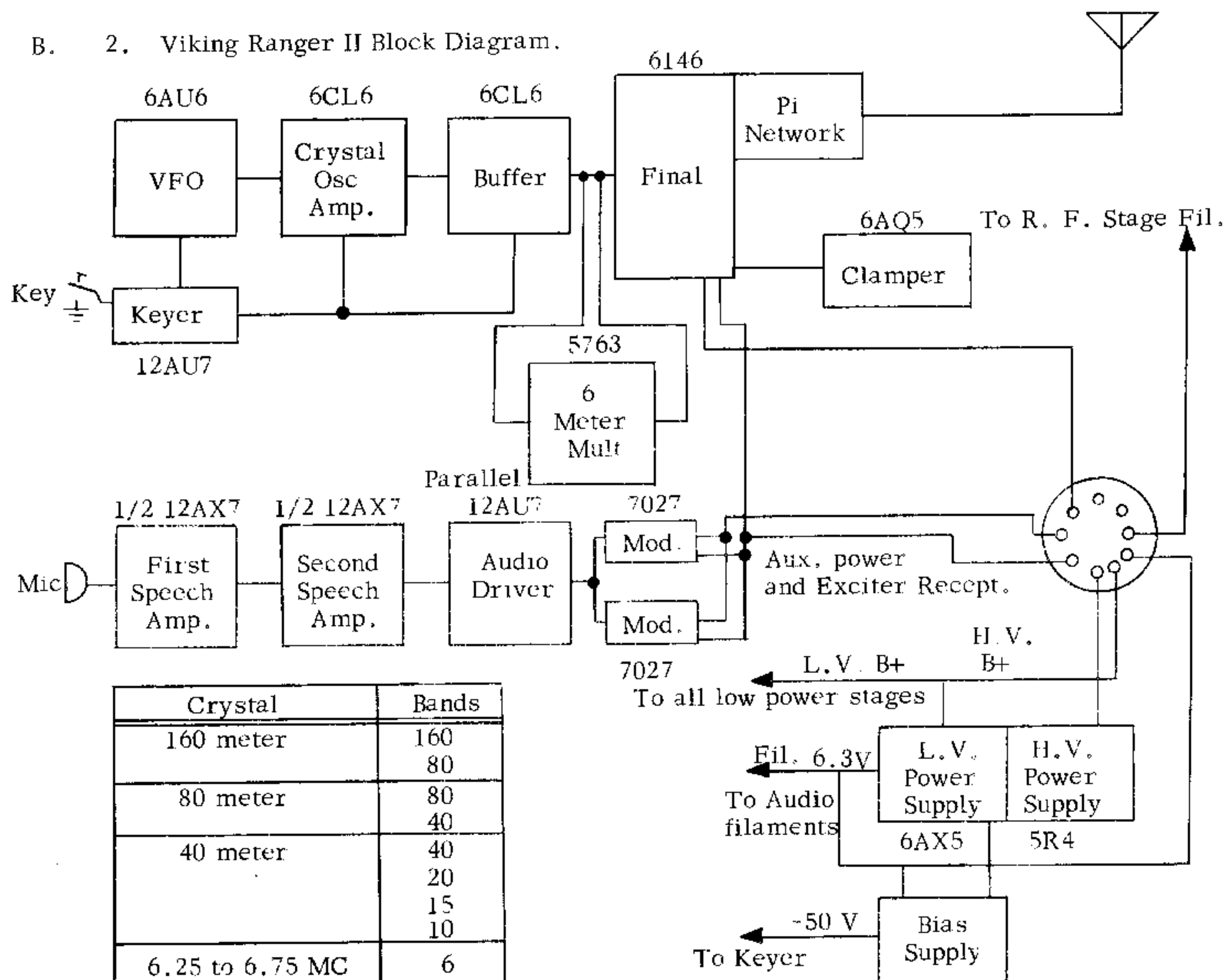


Chart shows crystals which may be used for full output on the amateur band covered by the "Ranger II".

3. Crystal Oscillator-Frequency Multiplier Stage. This stage employs a 6CL6 pentode tube of excellent RF characteristics. During VFO operation, this stage acts as an isolater-frequency multiplier, being broad tuned on all bands 160 through 20 meters. The plate circuit is ban switched automatically by the band switch. During crystal operation, this stage replaces the VFO and acts as a hot cathode crystal oscillator. During "CRYSTAL" operation this stage is keyed for CW transmission.
4. Crystal - VFO Switch. This switch selects either of two crystals that may be plugged into the crystal socket, J5, submounted at the front of the transmitter. In the XT1 position, a crystal connected between pin 3 and 5 of J5 is operative. In the XT2 position, the crystal connected between pins 1 and 7 is in use.
 - a. In the two crystal positions, XT1 and XT2, the switch disables the VFO and converts the first 6CL6 from an isolater-multiplier to a crystal oscillator and/or frequency multiplier.
 - b. In the "VFO" position, the switch places the VFO in operation and connects and converts the first 6CL6 from a crystal oscillator-multiplier to an isolater-frequency multiplier.

- B. 4. c. In the VFO "ZERO" position, it places the VFO in operation - plus whatever additional stages are required in each individual installation for a comfortable injection level into the receiver for positive zero beating purposes. A simple internal wiring change permits either the VFO alone or one or both of the 6CL6's to be energized in the VFO "ZERO" position.
5. Buffer Stage - 160-10 meters. This stage employs a 6CL6 RF pentode, employing a tuned high Q plate circuit, which is tuned to the same frequency as the final on all bands. This stage is protected against excitation failure by the cathode resistor R12. The buffer switch and coils are fully shielded to avoid any possible interaction. The drive control, R13, controls the screen voltage of this stage thus controlling its output and the final grid drive.
 6. Multiplier Stage - 6 meters. This stage employs a 5763 RF pentode with an untuned plate circuit, on the same frequency as the final. This stage is fed from the 6CL6 buffer and operates as a multiplier. It is switched in and out by SW3B.
 7. Final Amplifier. This stage employs a 6146 beam power amplifier with completely band switched pi-network plate circuit of Hi Q design. Great consideration and study has been made in the final coil assembly to maintain a constant Hi Q circuit for maximum degree of efficiency on all bands. The range of antenna impedances which may be matched on all bands is 50-500 ohms plus tuning out wide range of inductive or capacitive reactance. These output capacitance switching provisions are strategically located to avoid inductive loops coupling back to the previous stages. The range of antenna impedance which may be matched at frequencies above 7 mc extends, roughly, from 25 to 2000 ohms.
 8. Clamper Tube. This stage employs a 6AQ5 tube to furnish protection for the 6146 final amplifier in case of excitation failure. With excitation failure, the 6AQ5, connected in shunt with the screen dropping resistor, conducts and lowers the screen potential to approximately plate current cut off in the 6146. The screen of the clamper is connected to a voltage divider making the stage continue to conduct even at extremely low plate potential values. Under this cutoff condition, it will be noticed that 35 to 40 ma. current is indicated on the meter. This is not plate current of the 6146 - it is the plate current through the series screen dropping resistor, R15, drawn by the clamper tube.
 9. First and Second Speech Amplifiers. These stages employ a dual triode 12AX7 tube cascade connected with high gain design. Additional circuit design considerations control the response characteristics of these two stages. A two pin microphone jack is provided to accommodate the addition of a push-to-talk relay, if so desired.
 10. Audio Driver. This stage employs a 12AU7 dual triode, parallel connected and transformer connected to the modulator grids to furnish low impedance drive to the modulators.
 11. Modulator. This stage employs a pair of 7027 tubes, operating push-pull class AB. The modulators operate well within their ratings, and will deliver more than sufficient audio power for 100% amplitude modulation. Plate saturation limiting prevents large swings beyond full modulation thereby providing some limiting to reduce distortion and spurious output. The modulation transformer has a tertiary feed-back winding coupled to the grid of the audio power driver. This provides damping for improved regulation, stability and flat response and is particularly helpful in providing improved regulation for directly driving class B2 modulators when the Ranger II is used as an exciter. The secondary winding of the modulation transformer is center tapped to fill the requirements as an audio driver transformer when using the Ranger II as an exciter, working directly into the grids of a pair of Class B modulators. These leads are filtered and by-passed, and made available at the exciter-auxiliary plug at the rear of the chassis. By using one-half of the secondary winding, a nominal 500-600 ohm output can be obtained for driving large speakers used in paging or public address work. 33 watts of audio are available at the output of the modulators for any application required.

- B. 11. The frequency response range of the modulator section is flat within 3DB from 250 to 3000 CPS with very pronounced roll off above and below these frequencies. The audio quality is pleasing, yet retains the extra audio punch desirable for communications effectiveness.
12. Power Supplies. A dual voltage power supply, employing a 5R4GY HV rectifier tube and a 6AX5GT LV rectifier tube, furnishes the voltages required for the Viking Ranger II. Both the high voltages and low voltages power supplies employ choke input filtering for improved voltage regulation. The high voltage supply will deliver 500 - 525 VDC for final modulators, and the low voltage supply delivery 300 - 320 VDC for the RF exciter and speech low level stages. A 6AL5, V14, is used as a bias rectifier to supply bias voltage for the keying circuits and modulator bias.

If it is desired to power an external equipment, power may be taken from the exciter-auxiliary socket J-6. In addition to 33 watts of audio, there is available 6.3 VAC at 5.5 amperes for filament supply, 300 VDC at 50 ma. and 500 VDC at 210 ma. When the auxiliary plug is wired for external power, the complete RF section of the Ranger II (including filaments) is de-energized - the power supplies cannot supply the normal Ranger II full power requirements and external power simultaneously.

13. Keyer Control Circuitry. Grid-block keying is used on V1, V3 and V4. To avoid unnecessary interference with signals on adjacent channels, a wave-shaping filter consisting of R42-R43 and C89 is used on V3 and V4. To avoid chirp when the VFO is keyed, the keyer tube, V13, 12AU7, allows the VFO to start quickly - before V3 and V4 start conducting - and then continue operating until after V3 and V4 have stopped conducting.

The VFO keyer adjust control, R39, adjusts the "hold" time for VFO operation after the key is opened. This may be adjusted to cutoff the VFO between marks of keyed characters thus allowing rapid CW break-in operation by enabling the operator to be aware of incoming signals while he is keying the transmitter. When operating in this fashion, the slight arcing at the key may be noticed at a click in the local receiver only. It is not on the carrier. It can be reduced by installing a 2 1/2 millihenry RF choke (the ordinary receiver type) in each key lead as close to the key as possible, to reduce radiation of the spark energy by the key leads.

14. Operate Switch. This switch, located to the left of the VFO escutcheon, is a dual ceramic wafer rotary switch that provides the following functions:

- a. "Off" position. No voltages applied to any circuits.
- b. "Tune" position.

- (1) Turns on all filaments, dial light.
- (2) Turns on low and hi voltage power supplies.
- (3) Leaves screen voltage off modulators.
- (4) Disables final amplifier.

- c. "Phone" position.

- (1) Turns on HV indicator light.
- (2) Places final amp in operate condition.
- (3) Applies screen voltage to modulators.
- (4) Applies 117VAC to ANT relay jack.

- d. "Stand-by"

- (1) Turns off HV indicator light.
- (2) Disables final.
- (3) Turns off modulator screen source.

- B. 14. d. (4) Turns off 117VAC power to ANT relay jack
(5) Disables exciter.
- e. "CW"
- (1) Turns on HV indicator light.
 - (2) Short circuits modulation transformer secondary.
 - (3) Leaves screen source off modulators.
 - (4) Places final into operating condition.
 - (5) Applies 117VAC to ANT relay jack.
- f. The antenna relay jack in supplying 117VAC for external relay control is ideally suited to handle the switch functions of a larger final when the Viking Ranger II is used as an exciter.
- g. The problem of disabling the audio of a larger modulator in the "CW" position is partially solved since the grids would be short circuited when the Ranger II is in the "CW" position as an exciter. In addition, the secondary of the high-powered modulator should be shorted on "CW".

C. VIKING RANGER II TUNING PROCEDURE

1. NOTICE: The regulations of the Federal Communications Commission require a suitable license for operation of this equipment. Refer to publications of the Federal Communications Commission or the American Radio Relay League for the latest rules governing station and operator licensing.

Be sure to return the enclosed warranty registration card. This will register your transmitter at the factory. If it becomes necessary to write to the factory regarding your transmitter, refer to it by serial number.

The tuning procedure for the Viking Ranger II is identical on all bands of operation, 160 through 6 meters. Therefore, the discussion of tuning on one band will apply to all bands. Only the dial and switch settings will change when going from one band to another. A 40-75 watt light bulb should be used as a dummy load.

2. Set all knobs on the settings given below:
 - a. Operate switch, "OFF" position.
 - b. Drive control on "#5" position
 - c. Xtal - VFO SW on "VFO" position.
 - d. VFO - Pointer mid scale (7.215 mc).
 - e. Band switch, "40" meters.
 - f. Auxiliary coupling switch, position "5".
 - g. Coupling, position "5".
 - h. Meter switch, "GRID" position.
 - i. Audio, "0" position.
3. With the dummy load attached to J3, the transmitter connected to an adequate ground, and the AC plug in the 117VAC 60 CPS receptacle, all knobs set on the positions given above, tuning on 40 meters is accomplished as follows:
 - a. Turn the Operate switch to the "TUNE" position. The VFO dial should be illuminated. After a normal warm-up period, tune the "BUFFER" tuning knob for a peak indication of grid current on the meter. On 160 through 10 meters, adjust the DRIVE control for a reading of 2 ma. on the center scale which reads 0-10 ma. The 6 meter output does not drop off until grid current is below 1 ma.

NOTICE!!

DO NOT EXCEED 4 MA GRID CURRENT UNDER ANY
CIRCUMSTANCES OR 3 MA FOR PROLONGED PERIODS OF TIME

C. 3. b. Keyer Control Adjustment.

- (1) Couple the VFO output to a receiver (OPERATE switch in CW position).
 - (2) Close key and tune receiver to a VFO signal.
 - (3) Open key.
 - (4) Turn VFO keyer adjust control, R39, counter-clockwise until VFO signal starts.
 - (5) Turn VFO keyer adjust control, R39, clockwise just slightly beyond the point at which the VFO signal stops. Adjustment in the extreme clockwise position may cut off the VFO too soon and result in "squaring" or "sharpening" of the keying envelope at the break with attendant clicks.
- c. Turn the Meter switch to "PLATE" position.
- d. Turn the Operate switch to "PHONE" position and immediately tune the "FINAL" tuning knob to plate current dip (resonance) on the plate current meter.
- e. Increase the loading on the final by adjusting the Coupling controls. After changing the Coupling controls, retune the final to dip (plate resonance) on the meter. Successively adjust these controls (always dip final last) until a plate current of 130 ma. is read on the top meter scale, 0-200 ma. On CW (key down), the plate current should be 140-150 ma.
- f. This completes the tuning and loading of the transmitter on the 40 meter band, phone or CW operation. Operation on other bands merely requires switching the bandswitch to the desired band and adjusting of the Buffer, Final, and Coupling controls to obtain proper final amplifier loading. As previously listed, proper loading is obtained at a plate current of 140-150 ma. on CW and 130 ma. on Phone. The drive control should be adjusted to provide a grid current of 2 ma. on all bands when the final is fully loaded at the above values.
- g. The following table of dial settings gives the approximate dial settings for fully loaded operation into a 52 ohm non-inductive load:

Line Voltage 117.5 VAC Phone Operation 52 Ohm Coax Load				
Frequency Megacycles	Buffer Tuning	Final Tuning	Auxiliary Coupling	Coupling
29.700	72	55	6	1
28.900	57	50	6	3.5
21.450	69	44	6	1
21.000	62	40	6	1
14.350	50	43	6	1
14.000	42	42	6	0
7.300	64	42	4	6
7.000	52	30	4	4.5
4.000	90	71	3	4
3.500	50	40	3	1
2.000	68	77	6	5
1.750	20	20	4	4
52	22	84	7	5
50	15	82	7	5

It should be borne in mind that reactances at the feed point, or impedances different than the 52 ohms used in compiling the chart will cause a change in all the dial settings except Buffer tuning.

D. VIKING RANGER II PI-NETWORK TUNING AND HARMONIC SUPPRESSION

The pi-tuning/coupling network in the Viking Ranger II is designed to load the final amplifier into antenna resistances of nominally 50 to 500 ohms throughout the frequency range of the transmitter. In addition, it is capable of "tuning out" series antenna reactances up to several hundred ohms to complete a good match to most unbalanced antenna systems. The range of antenna impedances which may be matched by the pi-network at frequencies higher than 7.0 mcs. extends from roughly 25-2000 ohms.

When the transmitter is well grounded and properly tuned, the higher harmonic suppression is excellent, generally much better than with other conventional methods of antenna coupling. This should be of interest to amateurs afflicted with TVI or other high frequency interference problems.

1. Importance of Grounding.

To obtain proper tuning, coupling and harmonic suppression with any transmitter antenna coupling system, the part of the circuit designed to operate at RF ground potential must be at RF ground potential. A "room full of RF" is evidence that a high RF potential exists on something in or near the room. In many cases the source of RF is the transmitter's chassis and power cord. This condition is very undesirable for several reasons. The power cord is very closely coupled to the chassis by the electrostatic shields of the power transformers. Three objectionable factors which obviously affect the loading of the transmitter when poor grounds are involved are:

- a. The impedance that the output terminal of the transmitter looks into includes not only the true antenna to ground impedance as presented by the antenna feedline but also the transmitter chassis to ground impedance. This additional impedance in some cases will raise the apparent antenna impedance to such a high value that it cannot be loaded by the pi-network.
- b. Part of the transmitter's power is lost in the ground system due to radiation of the ground lead, power cord or cabinet. This power is quickly dissipated in surrounding objects and contributes nothing to effective radiated power except to distort the antenna's normal field pattern.
- c. It is conventional, in designing a transmitter, to by-pass harmonics or any possible source of stray high frequency currents to the chassis on the assumption the chassis will be kept as near ground potential as possible. When a high impedance is presented to these currents at the chassis, they are able to radiate to some extent rather than be passed harmlessly to ground.

2. How to Obtain a Good Ground:

What may appear to be a good ground at one frequency may prove to be a poor ground at another. A single ground lead may have "standing waves" on it due to its length. While it may seem difficult to obtain a good ground over a wide range of frequencies, it can be done and will be well worth the trouble when increased radiation efficiency, ease of antenna loading, and reduce TVI and BCI result. There is also reduced danger of damaging microphones, receivers, and other associated equipment with excessive RF fields.

Avoid using the power line, power line conduit or gas lines for RF grounding. Some suggestions which may help to obtain a good ground are

- a. Water pipes or metal building structural members are usually good sources of earth grounds.
- b. Use heavy conductors (#14 or larger) between the connection at the ground point and transmitter. Copper ribbon is excellent for this purpose.
- c. The use of several ground leads, each of a different length and selected at random may

- D. 2. c. be helpful in keeping grounding impedance low at the transmitter, even though the transmitter is some distance from a true earth ground. The possibility of obtaining an effective ground at any frequency throughout the transmitter's range is quite good. If at any one frequency one of the ground leads presents a low impedance at the chassis, the chassis is effectively grounded. By changing the length of one of the ground leads experimentally, a good ground can often be obtained at a frequency which has been troublesome. In bringing several leads to the transmitter, small closed loops near the transmitter or antenna feed line should be avoided. Induction fields will tend to raise the impedance of the ground leads.
- d. In cases where it is impossible to obtain a good earth ground, connecting the transmitter chassis to some system of conductors having a very low effective impedance to ground compared to the antenna impedance may be helpful. Usually this artificial "ground" takes the form of a system of radial wires spread horizontally on the floor, a gridwork of wires, or a large metal sheet on the floor below the transmitter. To be most effective, the minimum area covered by the metal conductors should be roughly equivalent to a square, the length of one side of which approaches a quarter wavelength. This system of "grounding" should be experimented with before committing the location of any permanent installation.
- e. A simple counterpoise made up of a single wire attached to the chassis may be helpful. On 10 meters, a length of 6 to 8 feet may be attached and the open end cut off 4 inches at a time until the chassis becomes "cold". The open wire may be allowed to drop along the floor although its open end will be somewhat "hot".
- f. A rough check on the effectiveness of the transmitter ground may be made by touching the chassis while watching the PA plate current and grid current with the transmitter operating into an antenna. A change in current upon touching the chassis is indicative of an ineffective ground. If a neon bulb, held between the fingers, can be ignited by touching it to the chassis, the RF present is excessive and is another indication of an ineffective ground. In cases where the transmitter is feeding a low impedance antenna, the test by touching the chassis is more reliable since 50 to 60 volts is required to ignite the neon lamp.

3. Loading Random Antennas with the Pi-Network:

With the transmitter chassis well grounded, correctly designed antenna systems having relatively "flat" unbalanced feeder systems, can easily be loaded by following the instructions already given, provided the antenna terminal impedances fall within the range of the pi-network. Feeding a balanced system with a feedline over a quarter of one wavelength long, may prove to be surprisingly successful if the transmitter chassis is held at ground potential. The transmission line between the transmitter and antenna will tend to assume a partial balance at the antenna. Some standing waves will result but may not be excessive. The Johnson Viking Matchbox, a universal all band, bandswitched antenna coupler will permit loading of the Viking Ranger II transmitter to any practical antenna system. In addition, it provides for the use of the Johnson 250-20 Low Pass Filter for increased harmonic suppression.

Antennas having random lengths, random feed points and various types of feed line will exhibit widely different resistance and reactance characteristics. It is well to remember that the feedline is a very important part of the system. A common example of the random antenna is a horizontal wire fed by a single wire feed line. The feed line in this case actually becomes part of the radiating system. An antenna of this type can, in most instances, be fed by the pi-network directly but there are critical dimensions where the antenna resistance can become either too high or too low to be matched by the pi-network.

Antennas with high terminal resistance or reactance can usually be recognized while loading the final stage of the Viking Ranger II. The final amplifier is normally loaded by reducing the output coupling capacitor, C9, in small steps, returning the amplifier to resonance each time. This results in an increase in PA plate current and is continued until full loading is achieved. If, however, a point is reached where decreasing the output coupling capacitor,

- D. 3. C9, does not result in a marked increase in PA plate current and the PA is not fully loaded, the antenna can be assumed to have a high resistance or reactance at this frequency.

Antennas with low terminal impedance (resistance and reactance both low) can usually be recognized by a noticeable lack of coupling capacitor effect in the range of settings normally used at the operating frequency. There will be little or no detuning evidenced as the coupling control is changed.

Several things can be tried in an effort to bring the antenna system into the tuning range of the pi-network:

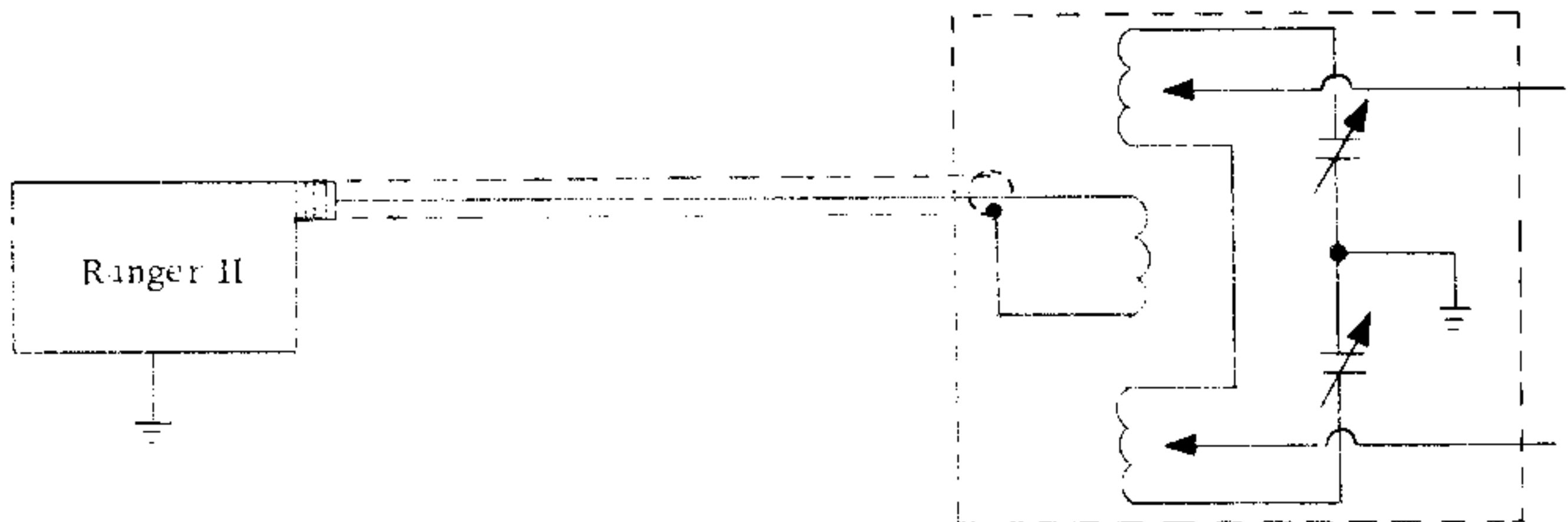
- a. Change the length of the feeder line between the antenna and transmitter experimentally $1/8$ to $1/4$ wavelength.
- b. Change the point of connection of the feedline to the antenna $1/8$ to $1/4$ wavelength.
- c. Change the antenna length $1/8$ to $1/4$ wavelength. Antennas shorter than $1/8$ wavelength (antenna and feeder) may be difficult to load. They present a high capacitive reactance to the transmitter output terminals. Effective antenna lengths in the vicinity of $1/2$ wavelength will in general exhibit characteristics of high resistance, high reactance, (inductive or capacitive), or both.
- d. "Load" the antenna feeder by placing an inductor or capacitor in series to cancel out the reactance of the antenna feeder. This may require considerable cut and try and will affect only the reactive component of the antenna impedance. However, it can prove useful in some cases.
- e. L type matching networks of inductance and capacitance may be used to aid impedance matching. Much discussion of this more elaborate method of bringing the antenna impedance within the range of the pi-network could be included, however, the few cases where it is necessary do not justify inclusion herein. Textbook and handbook discussions will be helpful if work along this line is pursued. There is danger of resonating the coupling capacitor of the pi-network when using an external coil. This should be watched as excessive voltage built up across the coupling capacitors can cause damage. Improper coupling or loading will take place under these conditions.

4. Dangers to be Avoided and Hints which may Further Aid in Harmonic and TVI Reduction.

- a. When loading high impedance antennas there is a temptation to "squeeze" the last watt into the antenna by opening the coupling capacitors as much as possible. Harmonic suppression is dependent, to a great extent, on the amount of coupling capacity in the circuit. It is wise to use as much coupling capacity as practical at all times. The proper amount of coupling when the antenna impedance is high, can be conveniently determined by holding a neon lamp against the antenna feeder. The coupling capacitor can then be opened until little increase in glow is noticed when the coupling capacitor and tuning controls are adjusted for maximum output. A decrease in coupling capacitance beyond this point may cause a higher plate current reading due to reduce plate circuit efficiency. Higher harmonic output will also result as the coupling capacity is reduced beyond the point where the output has leveled off. The random antenna system may present a more favorable impedance to harmonic output than the output on the fundamental frequency, hence it is well to use as much coupling capacity as is practical. It is well to remember that the amount of coupling capacitance needed is dependent on the operating frequency. For example, 2,000 micro microfarads at 3.5 mcs. corresponds to 160 micro microfarads at 28.0 mcs. These are the values necessary to couple resistive loads of approximately 50 ohms, at the frequencies mentioned.
- b. If the power line voltage is low or the high voltage rectifiers have low emission, the loaded plate current may not reach the normal value. This condition should not be confused with the inability of the pi-network to load an antenna system.

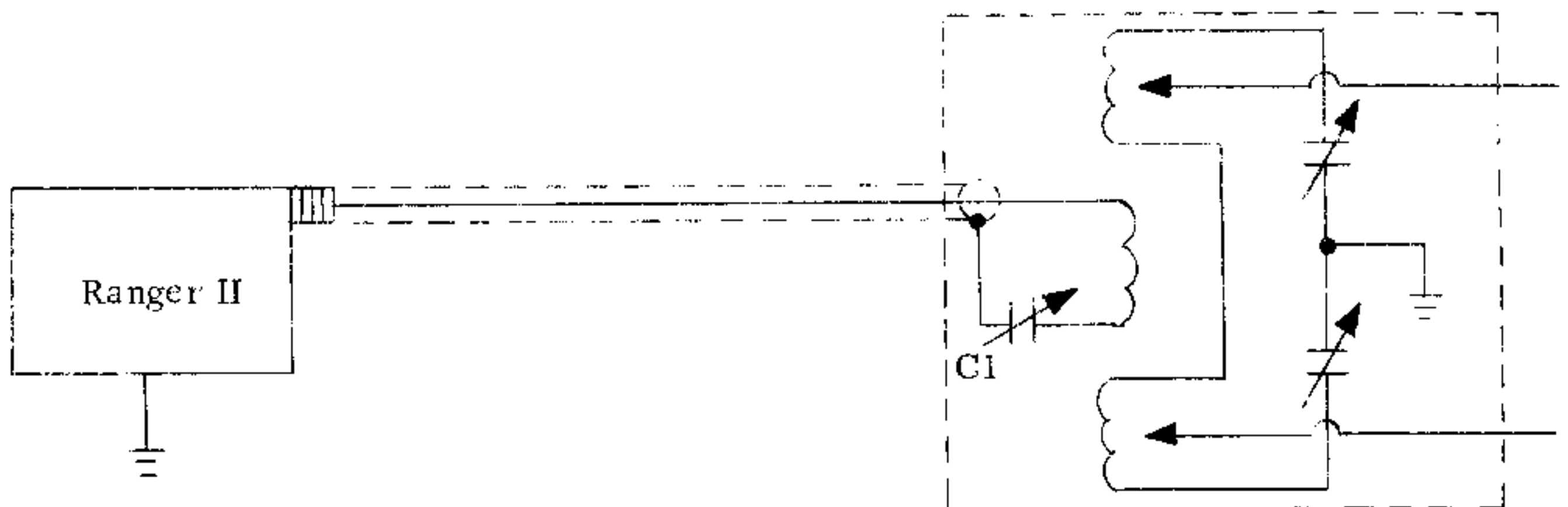
D. 5. Coupling to Balanced Antennas:

Balanced antennas such as center fed "Zepps", beams and folded dipoles normally use a two wire transmission line and should have equal voltages, 180 degrees out of phase, applied to each feedline terminal. Since the output of the Viking Ranger is single ended (unbalanced), a coupler is required for balanced antenna systems. The Johnson Matchbox, a universal all band, bandswitched antenna coupler will permit loading of the Viking Ranger II to any practical antenna system. In addition, it provides for the use of the Johnson 250-20 Low Pass Filter for increased harmonic suppression. A simple coupler for this purpose is shown below. The tank circuit is resonant at the operating frequency and can be excited by a coaxial line and coupling link. Line impedance is not critical although 52 ohm line will be most desirable if a Johnson Low Pass Filter is to be used.



Feedpoint impedance of the coupler is adjusted by means of the inductor taps. Tap adjustment is unnecessary with the Johnson Matchbox. Final amplifier loading is adjusted with the transmitter output coupling controls.

Tuning of the coupler can be made quite broad by making the L/C ratio as high as possible (low "Q") while still permitting the desired loading. Inductive reactance of the coupling link may make it impossible to reduce the SWR of the coaxial line to or below 1 1/2 to 1. If so, the link circuit may be made series resonant by adding capacitor C1 as shown below:

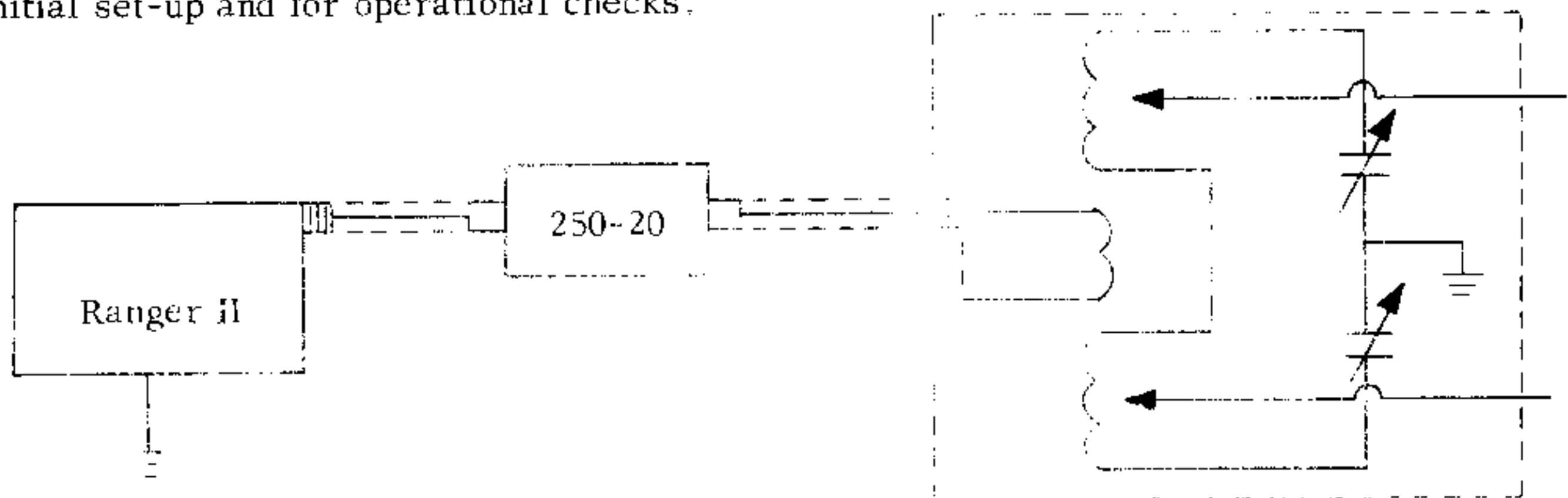


The above problem is non-existent with the Matchbox.

6. Use of Low Pass Filters:

Depending upon how it is tuned, 2nd harmonic attenuation of the Viking Ranger II amplifier can be as high as 30 db. Since this will permit operation in many locations without television interference, the Johnson 250-20 Low Pass Filter is not an integral component of the Viking Ranger II but is available as an optional accessory. This filter will provide an additional 75 db or more harmonic attenuation with insertion loss of less than .25 db. The filter's

- D. 6. characteristic impedance is 52 ohms and has a power rating of 1 KW. The low pass filter may be inserted in the coaxial line between the transmitter and the antenna coupler. Coaxial connectors are used at the transmitter and at both ends of the low pass filter to preserve the shielding provided by the coaxial line. It is preferable that the standing wave ratio on the coaxial line be maintained at 2 to 1 or less, therefore the impedance of the line between the Viking Ranger II and the coupling link should be the same as the characteristic impedance of the filter. (The Johnson 250-20 Low Pass Filter and Johnson Matchbox are 52 ohm impedance.) The section of coaxial line between the transmitter and the low pass filter should be as short as possible and electrical quarter waves should be avoided. A Directional Coupler, such as the Johnson 250-37, for measuring SWR will prove invaluable for both initial set-up and for operational checks.



An end fed half-wave antenna may present loading problems, both from the standpoint that its impedance is higher than can be matched by the pi network amplifier of the Viking Ranger II, or that the low output coupling capacitance used reduces inherent harmonic attenuation below tolerable values. Therefore, the use of a half-wave antenna may create TVI problems, while other antennas prove perfectly satisfactory. In these cases, it is recommended that the Johnson Matchbox be used.

E. VFO CALIBRATION PROCEDURE

1. Signal Generator, Receiver, and VFO Setup for the Viking Ranger II VFO Calibration.

- a. The accuracy of the Viking Ranger II VFO will be no better than that of the signal generator used to calibrate it. To fully utilize the stability and calibration capabilities of the VFO, the frequency standard used to calibrate it should have an accuracy of .005% or better. Most crystal standards or crystal calibrated variable frequency standards are satisfactory for normal calibration purposes. A moderate signal output capable of being easily detected by the receiver to be used for zero beat indication, is necessary at the following frequencies.

F1a Any given frequency (preferably a VFO low frequency scale mark frequency) between 1.75 to 1.78 mc or any of the first eight harmonics of 1.75 to 1.78 mc in the range of the receiver. 1.76, 3.52, 5.28, 7.04 and 8.80 mc are good calibrating frequencies.

F2a Any given frequency (preferably a VFO low frequency scale mark frequency) between 1.96 and 2.00 mc or any of the first eight harmonics of 1.96 to 2.00 mc in the range of the receiver. 1.97, 3.94, 5.91, 7.88, and 9.85 mc are good calibrating frequencies.

F3a Any given frequency (preferably a VFO high frequency scale mark frequency) between 7.00 and 7.07 mc or any of the first four harmonics of 7.00 to 7.07 mc in the range of the receiver. 7.03, 14.06, 21.09, and 28.12 mc are good calibrating frequencies.

- E. 1. a. F4a Any given frequency (preferably a VFO high frequency scale mark frequency) between 7.35 and 7.425 or any of the first four harmonics of 7.35 and 7.425 mc. 7.40, 14.800, 22.2, and 29.6 mc are good calibrating frequencies.
- b. Warm up the signal generator for at least 1/2 hour or as long as suggested by the signal generator instructions before using it for VFO calibration.
- c. Set up a receiver capable of detecting each of the frequencies chosen in 1a. Attach antenna leads to the receiver input and the signal generator output and bring the leads together until signal generator output can be picked up by the receiver. Separate and shorten the leads as found necessary to keep the receiver from blocking due to excessive signal input. Allow the receiver to warm-up for about 1/2 hour to stabilize the local oscillator, and log dial settings for frequencies F1a, F2a, F3a and F4a. The beat frequency oscillator in the receiver may be used to log and compare the signal generator and VFO frequencies but it is desirable to obtain the final zero beat indications between the VFO and the signal generator signals without the beat frequency oscillator. Avoid setting the receiver on or logging image frequencies.
- d. Warm-up the Viking Ranger II in the "Tune" position for 1/2 hour. Turn the band-switch to the 160 or 80 position. Turn the VFO dial pointer to the frequency F1, between 1.75 and 1.78 mc., chosen as the low 160 meter calibrating point, and find it or its harmonic (near F1a) on the receiver. Repeat the same procedure at the high 160 meter calibrating point and the 40 meter high and low points after moving the Band-switch to the 40 meter position.

2. 160, 80 Meter VFO Scale Calibration

- a. Set the Bandswitch on the 160 or 80 meter position and the dial at F2, the frequency between 1.96 and 2.00 mc chosen for the high 160 meter calibrating point. Set the signal generator and the receiver at F2a. Adjust the "160 hi" trimmer at the top of the VFO (refer to Figure 1) until the VFO zero beats with the signal generator.
- b. Turn the VFO to F1, the receiver to F1a, the signal generator to F1a, and adjust the "160 lo" padder until the VFO zero beats with the signal generator.
- c. Repeat the "160 hi" and "160 lo" adjustment, zero beating the signal generator and VFO as accurately as the ability to reset the two units warrants.

3. 40, 15, 10 Meter VFO Scale Calibration

- a. Set the Bandswitch on the 40 or 20 meter position and the dial pointer at F4 on high frequency dial scale, the frequency between 7.35 and 7.425 mc chosen for the high 40 meter calibration. Set the signal generator and the receiver at F4a. Adjust the "40 hi" trimmer at the top of the VFO until the VFO zero beats with the signal generator.
- b. Turn the VFO to F3, the frequency between 7.00 and 7.07 mc chosen for the low 40 meter calibration, the receiver to F3a, the signal generator to F3a, and adjust the "40 lo" padder until the VFO zero beats with the signal generator.
- c. Repeat the "40 hi" and "40 lo" adjustment, zero beating the VFO as accurately as the ability to reset the two units warrants.

4. 6 Meter Calibration

- a. The 6 meter band VFO output is from 6.25 to 6.5 mc. A given frequency, F5a, in the range 6.25 to 6.5 mc or any of the first four harmonics of the 6.25 to 6.5 mc range may be used to calibrate the 6 meter range. Turn the bandswitch to the 6 meter position and set the dial pointer on the standard frequency F5 or the harmonic of the standard signal which falls in the 6 meter band. Set the receiver to the 6 meter range or a sub-harmonic and detect the standard signal frequency. Adjust the "6 meter" trimmer until the VFO zero beats with the standard frequency.
- b. Recheck the 40 or 20 calibration after the 6 meter adjustment. There is little likelihood that further re-adjustments are necessary unless a large change was required in the "6 meter" setting.

E. 5. VFO Calibration Using the Transmitter Crystal Oscillator or Other Standard Signal Sources.

- a. Crystals of known frequency and accuracy in the frequency ranges F1a, F2a, F3a and F4a (designated in paragraph E1a) can be used in the transmitter crystal oscillator to provide standard frequency signals for the VFO calibration. The stability of the receiver local oscillator and beat frequency oscillator must be nominally good as the technique of beating the receiver BFO to the crystal and then beating the VFO signal to the receiver will be used. The receiver thus "remembers" the crystal frequency. Reduce the coupling of the receiver antenna to the minimum usable amount to avoid "pulling" the receiver local oscillator.
- b. An example of calibrating the VFO using actual crystal values may be helpful. Assume that the following crystals have been found as part of the amateur station equipment: 7060 kc, 6250 kc, 3690 kc, and 1980 kc. The dial calibration points then become:

$$F1 = \frac{7,060}{4} = 1,765 \text{ mc.}$$

$$F2 = 1,980 \times 1 = 1,980 \text{ mc.}$$

$$F3 = 7,060 \times 1 = 7,060 \text{ mc.}$$

$$F4 = 3,690 \times 2 = 7,380 \text{ mc.}$$

The receiver setting and VFO harmonic which may be used for each respective dial calibration frequency becomes:

$$F1a = 7,060 \times 1 = 7,060 \text{ mc}$$

$$F2a = 1,980 \times 4 = 7,920 \text{ mc}$$

$$F3a = 7,060 \times 1 = 7,060 \text{ mc}$$

$$F4a = 3,690 \times 2 = 7,380 \text{ mc}$$

Proceed as follows:

- (1) Place the 1,980 mc. crystal in the XT1 position and the 7,060 mc. crystal in the XT2 position.
- (2) Set the Bandswitch on 160 or 80 meters, the VFO dial pointer on the 1,980 mc. mark, the Crystal VFO switch to the 1,980 mc. crystal position (XT1). Tune the receiver to zero beat the BFO with the crystal. Turn the Crystal VFO switch to "VFO" and adjust the "160 hi" trimmer to zero beat the BFO.
- (3) Set the VFO pointer on the 1,765 mc. mark and the Crystal VFO switch to the 7,060 mc. position (XT2). Tune the receiver to zero beat the VFO with the crystal. Turn the Crystal VFO switch to "VFO" and adjust the "160 lo" padder to zero beat the BFO. Repeat steps (2) and (3) to minimize adjustment interaction.
- (4) Remove the 1,980 mc. crystal and replace it with the 3,690 mc. crystal in the XT1 position.
- (5) Set the Bandswitch on 40 meters, the VFO dial pointer to 7,380 mc., and the Crystal VFO switch to XT1. Tune the receiver to zero beat the BFO with the crystal. Turn the Crystal-VFO switch to VFO and adjust the "40 hi" trimmer to zero beat the BFO.
- (6) Set the VFO pointer on 7,060 mc. and the Crystal VFO switch to XT2. Tune the receiver to zero beat the BFO with the crystal. Turn the Crystal VFO switch to VFO and adjust the "40 lo" padder to zero beat the BFO. Repeat steps (5) and (6) to minimize adjustment interaction.
- (7) The 6 meter band setting may be made with a crystal which will place a harmonic signal in the 6 meter band. Set the Bandswitch on 6 meters, the Crystal VFO switch to the crystal (assume 6250 kc is available) position. Zero beat the receiver BFO to 50 mc (the 8th harmonic of 6250 kc). Turn the Crystal-VFO switch to "VFO" and adjust the "6M" trimmer to zero beat the VFO to the receiver BFO.

E. 5. c. The user may think of several sources of standard signals other than those mentioned. In each case, the accuracy of the source should be known before using it. Many combinations of harmonics can be found and no attempt has been made to cover all of them in this discussion. Other signal sources which may be used but have not been discussed are:

- (1) The signal of another amateur station whose frequency has been determined by a standard.
- (2) The harmonics of a signal generator whose output signal has been zero beat with a broadcast station.
- (3) Signals of W W V discussed in the next section.

The VFO user must adapt his techniques to the signal source he has available.

d. Band edge crystals or crystals near the usual operating frequencies of the amateur stations are always valuable for occasional monitoring of the VFO signals. They may be used in a separate oscillator circuit or the crystal oscillator stage of the transmitter.

6. VFO calibration using the W W V 10 mc Signal. This calibration is not recommended if other standard signal sources are available. It will be noted that most calibration points are on the ends of the bands and the 40, 20, 15 or 10 meter band high scale calibration includes the tracking error of the low frequency 160, 80, or 40 meter band. The receiver, the receiver BFO, and the VFO should be warmed up for 1/2 hour before calibrating.

a. 160, 80 Meter Calibration.

- (1) Zero beat the receiver BFO to the 10 mc W W V signal.
- (2) Set VFO dial pointer to 2.00 mc and the Bandswitch on 160 or 80 meters.
- (3) Adjust the "160 hi" VFO trimmer until the fifth harmonic of the VFO is zero beat with the receiver BFO.
- (4) Leaving the VFO at this setting, zero beat the receiver BFO with the seventh harmonic of the VFO (14 mc).
- (5) Turn the VFO to 1.75 mc and adjust the "160 lo" VFO padder to zero beat the eighth harmonic of the VFO with the receiver BFO.
- (6) Adjust both ends of the 160 or 80 meter bands to zero beat the eighth and seventh harmonics of the VFO with the receiver BFO as necessary.

b. 40, 20, 15, 10 Meter Calibration.

- (1) Set the VFO dial at the 1.85 mc mark and zero beat the receiver BFO to the eighth harmonic of the VFO frequency at 14.8 mc.
- (2) Set the Bandswitch to 40 meters and the dial pointer to the 7.40 mc mark. Zero beat the second harmonic of the VFO to the 14.8 mc receiver setting by adjusting the "40 hi" trimmer.
- (3) Set the Bandswitch and dial pointer for 1.75 mc output again and zero beat the receiver BFO at 14 mc. Set the Bandswitch and dial at the 40 band low frequency end (7.00 mc) and adjust the "40 lo" padder to zero beat the VFO second harmonic with the receiver 14.0 mc BFO setting.

c. 6 Meter Calibration.

- (1) Set the Bandswitch and dial for 6.250 mc output.
- (2) Tune the receiver to 50 mc and zero beat the receiver BFO to the 8th harmonic of the VFO.
- (3) Set the Bandswitch on 6 and the dial pointer on 50.0 mc. Adjust the "6 meter" trimmer to zero beat the eighth harmonic of the VFO to the receiver BFO setting.

E. 7. Things to Look for if the VFO Frequency Cannot be Adjusted to the Dial Markings Due to Apparent Lack of Trimmer or Padder Range.

- a. Check to make certain the frequency standard used is accurate (crystals used in amateur service are often found to differ from their marked frequency due to holder conditions, oscillator circuit loading, or non-critical original calibration).
 - b. Make certain image frequencies are not being mistaken for desired frequencies in the receiver.
 - c. If, after checking the frequency standard and receiver settings, the VFO frequency cannot be adjusted to chosen dial marks, adjust the trimmers and padders to bring the VFO frequency as close as possible to the dial mark frequencies. Remove the VFO side cover and recheck the dial location relative to the tuning condenser shaft. The VFO tuning condenser should be exactly meshed (not necessarily the stop position) when the dial pointer is at the left horizontal position. If the dial requires relocating, try calibrating the VFO scale again, as directed in previous instructions, with the side covers on.
8. Temperature compensation adjustment C100, which is a differential variable capacitor, together with C98 and C99 form an adjustable temperature compensating network. C100 was set at mid-value during assembly and the temperature compensation will be approximately right. However, if more exact temperature compensation is needed, the following step by step procedure should be followed.

CAUTION: The signal source used to compare the Ranger II VFO must be very stable. A 100KC crystal calibrator is suitable. Do not rely on a receiver's frequency stability, as many receivers drift more than this VFO. The receiver will be used only as a mixer and as such will not affect the accuracy of the measurements. Warm up all equipment a minimum of one hour.

Procedure:

- (1) Warm up a receiver capable of tuning 7.2 megacycles.
- (2) Warm up the 100KC crystal calibrator.
- (3) Tune the receiver to 7.2 megacycles. This is best done by turning on the receiver BFO, and tuning the receiver to zero beat against the 100KC crystal calibrator at 7.2 MC and then turning off the BFO.
- (4) Apply power to the Ranger II, turn the bandswitch to the 40 meter position.
- (5) Connect a short antenna lead to the receiver and couple the receiver to the Ranger II by laying the antenna lead near the 6CL6 stages. Do not overload the receiver (if the 100KC crystal calibrator is not built into the receiver, it will have to be coupled to the receiver input also).
- (6) Allow the Ranger II to warm up for 30 minutes.
- (7) Carefully zero beat the VFO signal against the crystal calibrator signal.
- (8) Allow about 10 minutes (or however much time is necessary) until the VFO has drifted an audible number of cycles. Then carefully re-zero beat the VFO, being careful to note which direction the VFO must be turned to re-zero.
- (9) If the frequency of the VFO increased (the VFO signal was lowered in frequency to re-zero) C100 should be turned slightly clockwise (less negative temperature coefficient) then readjust 40 meter high trimmer to reset frequency if necessary.
- (10) If the frequency of the VFO decreased (the VFO signal was raised to frequency to re-zero) C100 should be turned slightly counter-clockwise (more negative temperature coefficient).
- (11) Repeat steps 8 and 9 as necessary.
- (12) Typical stability after a 30 minute warm-up, drift of less than .001% per hour is typical.

F. EXCITER OPERATION, FOR ZEROING, PUSH-TO-TALK

1. Instructions for Using the Johnson Viking Ranger II as an Exciter and Audio Driver.

To utilize the Ranger II as an exciter rather than as a transmitter, no internal changes are required, as can be seen by studying the schematic diagram at the rear of this manual. The exciter-auxiliary power socket, J-6, a 9-pin octal style socket, makes all the internal sources of operating potentials readily accessible with all outgoing leads filtered and by-passed for TVI suppression.

The audio output of the modulator section can be coupled directly into the grids of a high powered modulator by observing the following steps in making connections to the exciter-auxiliary power plug, P3:

- a. Using shielded leads, and not exceeding runs of 26 feet in length, connect the two modulator grids to pins 2 and 3 of the plug. Shields should be connected to pin 9. If runs exceeding 25 feet in length are anticipated, connections can be made to terminals 1 and 2, and a 500-600 ohm to push-pull grids transformer used at the modulator. This transformer must be a driver type if the modulator is operated class AB2 or class B2.
- b. Connect the bias supply to pin 1 of the plug to furnish bias on the modulator grids, using a shielded lead. The shield should be connected to terminal 9 which is chassis ground at the Ranger II exciter. If a separate line of grid transformer is used, the bias must be supplied through the center-tap of this transformer.
- c. Connect jumpers from pins 7 to 8 (for filaments) and from pins 5 to 6 (for H.V. to final).
- d. RF excitation can be obtained at the coax connector J3 at a level depending upon the requirements of the external amplifier. Continuous control is easily accomplished by merely varying the "DRIVE" control on the Viking Ranger II. A maximum of 40 to 45 watts output is available on Phone and 45-50 watts on CW under normal line voltage conditions. When connected to the external amplifier which is being driven, the Ranger II should be loaded to provide slightly more than the required grid drive on the external amplifier (the Ranger II should be dipped, of course, at this loading point with the Ranger II grid current at the normal value of 2 ma.) The drive control on the Ranger II should then be turned counter-clockwise (toward zero, thus reducing the power output of the Ranger II) to obtain the specified drive to the external amplifier.

2. VFO Injection in Receiver - Control of

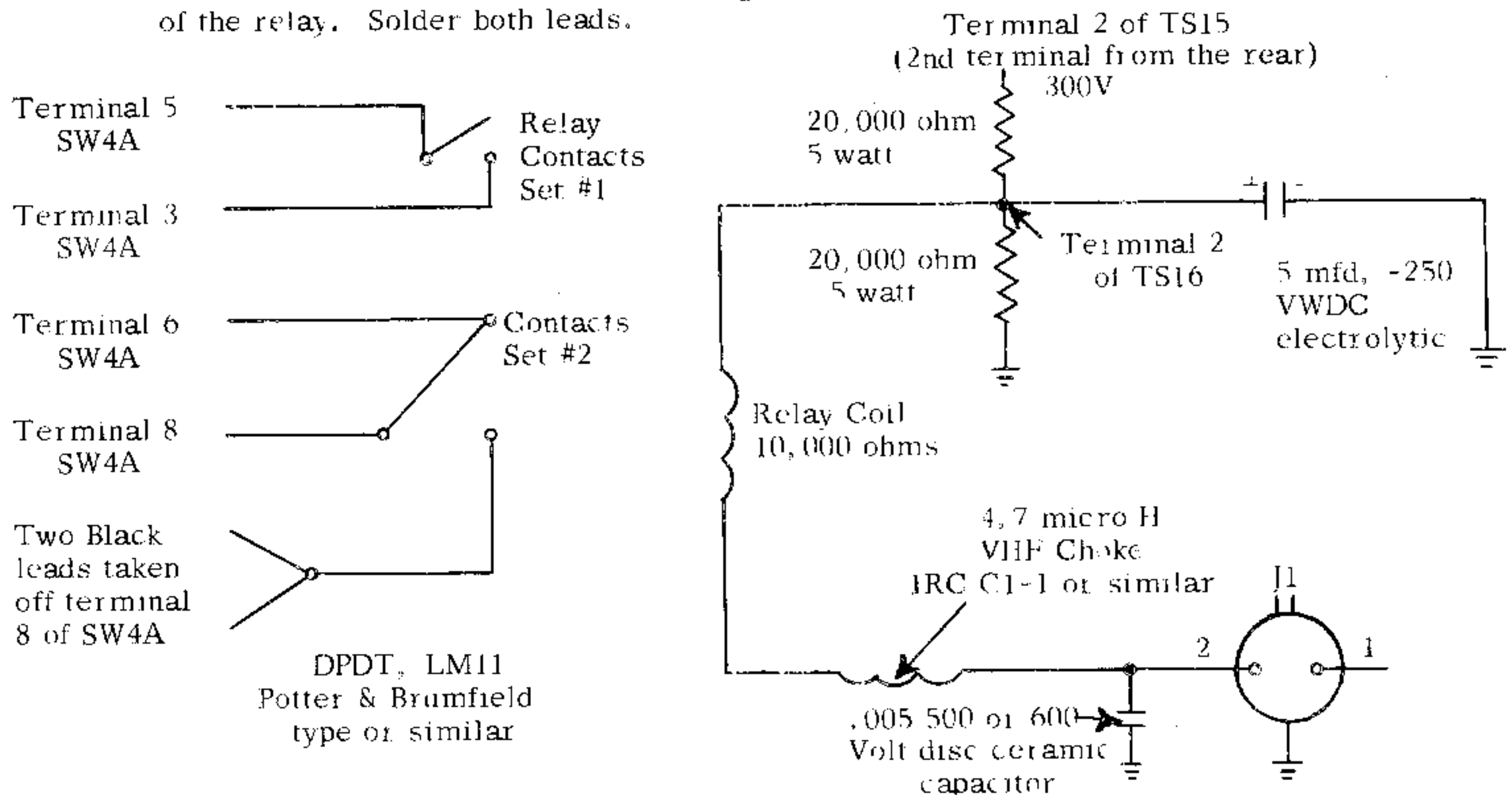
As normally wired, the "CRYSTAL-VFO" switch "ZERO" position energizes the VFO and the first 6CL6, V3, when it is desired to zero beat an incoming received signal. Dependent upon the individual station setup (relative location of transmitter, receiver, transmission lines, etc.), it may be desired to either increase or decrease the injection level which is accomplished as follows:

- a. To increase injection, place a short jumper between clips 11 (normally not connected) and 12 of the forward wafer of the Operate switch, SW4A. These clips are accessible from the top of the chassis. This change adds the buffer 6CL6, V4, in the "ZERO" position.
- b. To decrease injection, remove the jumper to terminal 4 of SW2.

3. Push-to-Talk Addition (see following page for schematic diagram)

- a. Procure the relay, two 20,000 ohm resistors, a 5 mfd 250 WVDC condenser, a VHF choke and the .005 bypass capacitor as indicated on the schematic diagram.
- b. Attach the relay on the top side of the chassis between V3 and V4, shown in Figure 1, taking care to clear parts when drilling the mounting holes and a clearance hole for two insulated leads, in the chassis. Install a solder terminal at one of the mounting feet, on the topside of the chassis.

- F. 3. c. Connect one of the 20,000 ohm resistors between terminal 2 (the second from the rear) of terminal strip TS15 (to left of socket XV12) and terminal 2 (the second from the rear) of terminal strip TS16 (to left of socket XV11). Connect the other 20,000 ohm resistor between terminal 2 of TS16 and a solder terminal (must be installed) at the high voltage filter condenser mounting stud (ground). Connect an insulated wire to terminal 2 of TS16 extending it through the chassis to one of the coil terminals of the relay. Solder all connections made thus far.
- d. Connect and solder a lead from terminal 2 of TS16 to the relay coil. Connect the other terminal of the relay coil to a long insulated lead, capable of reaching the terminal strip, TS26, near the microphone jack, as it is trained along the harness to TS26. Train the lead, pass it through one of the microphone-key shield notches and connect it to the middle terminal of TS26. The shield may be removed to facilitate this step and the one following.
- e. Connect a .005 mfd. disc ceramic capacitor between the ground solder terminal near the microphone jack, J1, and terminal 2 on J1. Connect the VHF choke between terminal 2 of J1 and the middle terminal of TS26. Solder all solder terminal and terminal connections made in this step. Replace the shield.
- f. Remove the black jumper from terminals 3 and 5 on the "Operate" switch SW4A (front wafer). Connect and solder a twisted pair of leads between these terminals and one set of normally open contacts on the relay. The blue-orange lead should remain connected to terminal 5.
- g. Remove the two black leads from terminal 8 of SW4A. Connect one lead of another twisted pair to terminal 8 and the other lead to the two black leads just removed from terminal 8. A single terminal can be installed beside SW4 for this connection or the connection can be soldered and taped neatly. Connect the other end of the twisted pair across the other set of normally open contacts on the relay. Solder all of the last twisted pair connections. Connect a lead from terminal 6 of SW4A to the normally closed contacts on the same side of the relay.
- h. Connect the positive end of the 5 mfd condenser to the coil terminal previously connected to terminal 2 of TS16 and the negative lead to the solder terminal at the foot of the relay. Solder both leads.



Schematic-Wired Connections of the Push to Talk Circuit

- i. The microphone plug and push-to-talk circuit should be wired to correspond to the J1 connections.
- j. For operation, the Operate switch should be placed in the Phone position thus permitting control by the microphone push-to-talk switch.

TABLE 1

VIKING RANGER II TRANSMITTER SOCKET VOLTAGES

[illegible]

NOTES: 1. Voltage at VFO terminals TS-34

2. Screen Voltage subject to change with drive control.
3. Grid Voltage measured at pin 7 of XV-6.
4. #2 voltage subject to change with Keyer adjustment.
5. Operate switch in CW pos. Key up.
6. Not significant with 20K ohms/v. meter.

TABLE 2

VIKING RANGER II TRANSMITTER SOCKET RESISTANCE MEASUREMENTS

Line Voltage Operate Switch Bandswitch	None Phone 6 Meters	Resistance Tolerance, Plus or Minus 20%								
		Audio Gain Full Clockwise								
		Resistance Measured from Chassis Ground								
TUBE SOCKET	1	2	3	TUBE SOCKET PIN NUMBERS				7	8	9
				4	5	6				
XV-1---6AU6				(See Note 1)						
XV-2---OA2				(See Note 1)						
XV-3---6CL6	3	200K	100K	0	0	25K	0	0	100K	200K
XV-4---6CL6	400	150K	15K	0	0	25K	0	0	15K	150K
XV-5---6146	0	0	50K	0	22K	0	0	0	0	
XV-6---6AQ5	25K	0	0	0	50K	25K	22K	0		
XV-7---12AX7	700K	1 Meg.	4.7K	0	0	500K	300K	0	4.7K	0
XV-8---12AU7	25K	140K	820	0	0	25K	140K	0	820	0
XV-9---7027	9.5K	0	20K	9.5K	5K	5K	0	0	.5	
XV-10--7027	9.5K	0	20K	9.5K	5K	5K	0	0	.5	
XV-11--5R4GY	20K	20K	N.C.	100	N.C.	100	N.C.	0	20K	
XV-12--6AX5	N.C.	0	60	N.C.	60	N.C.	0	0	25K	
XV-13--6AL5	17	15K	0	0	17	N.C.	15K	0		
XV-14--12AU7	22K	1 Meg.	25K	0	0	25K	0	0	25K	0
XV-15--5763	25K	INF.	0	0	0	30K	470	0	90K	90K
XV-14-NOTE 2	22K	1 Meg.	25K	0	0	25K	800K	0	25K	0

NOTES: 1. Resistances at VFO terminals TS-34
2. Operate Switch in CW Position, Key Up.

1--25K
2--0
3--200K
4--7
5--.5
6--3
7--100K
8--20K

G. 1. c. Transformer and Choke Winding Resistances and Open Circuit Voltages.

<u>Unit</u>	<u>Winding</u>	<u>Color of Leads</u>	<u>Resistance</u>	<u>Open Ckt. A.C. Voltage</u>
T1, Power Transformer	117 volt pri.	black to black	0.9 ohms	
	H.V. sec.	red to yel-red	95	750 volts
		yel-red to other red	95	750
	L.V. sec.	blue to yel-red	55	460
		yel-red to other blue	55	460
	Fil.	green to green	0.7	6.9
	H.V. rect.	yel to yel	0.5	5.6
	Bias rect.	brown to brown		150
T2, Mod. Transformer	Pri.	red to brown	100	
		red to blue	100	
	Sec.	green to black	28	
		black to yel	28	
	tertiary	white to black-white	7	
T3, Audio Driver Transformer	Pri.	red to blue	270	
	Sec.	yel to black	135	
		black to green	135	
L33	H.V. choke	black to black	95	
L34	L.V. choke	black to black	280	

Individual transmitters may vary somewhat from the values listed above, but no more than 10 to 20%.

2. Trouble Shooting

- Be careful while making High Voltage Measurements. Do not take chances.
- Never depend on Bleeder Resistors to discharge Condensers. When turning equipment off, discharge each filter capacitor with a screwdriver equipped with a well insulated handle.
- All Power Supplies must be off and discharged when making ohmmeter measurements to prevent damage to the ohmmeter.
- Schematics, photographs, and charts aid greatly in trouble shooting and are furnished in this section for reference. Particularly useful will be the typical operating voltages, current reading, and the resistance measurements. Use these charts and listings to save time in locating trouble.

- It is almost impossible to anticipate all troubles, operating errors, or component failures in the following listing. It is attempted, however, to list possible combinations that will aid in correcting trouble normally encountered in transmitter construction and operation.

- G. 3. NEVER REPLACE THE FUSE, FUSTRON OR SLO-BLO FUSE, WITH VALUES LARGER THAN THOSE SUPPLIED AND RECOMMENDED.
4. Fuse blows when Operate switch is turned to the "TUNE" position.
 - a. In order to help determine the location of the short, remove the exciter-auxiliary power plug, P3. This will disable the filaments and low voltage supply to the VFO and RF section. If the short condition prevails, it will be found in the modulator section of power supplies.
 - b. With the exciter-auxiliary plug, P3, removed, power off, measure the resistance from pin 4 of the exciter-auxiliary power socket J-6 to ground. The reading should be 25,000 ohms. If reading is correct, the L.V. Supply is probably alright.
 - c. Check all tubes for internal shorts between plate and other elements.
 - d. Measure primary resistance of power transformer T1.
 5. Fuse blows when "Operate" switch is turned to "PHONE" position.
 - a. With the exciter-auxiliary plug, P3, remove and power off, measure between pins 5 and 6 of exciter-auxiliary socket J-6, and ground. This will check for shorts in the final plate, screen and clamper circuit. Resistance should be many megohms.
 - b. Measure secondary resistances of the power transformer T1. CHECK THE H.V. CAPACITOR, C77.
 - c. With the exciter-auxiliary power plug P3 removed, measure from pins 1, 2 and 3 of XV 13 to ground. Readings should be many megohms.
 - d. Check all tube sockets for evidence of shorts. Any "sweating" tendency on the part of the disc ceramics may indicate an internal high resistance leakage.
 6. RF Exciter section. "TUNE" position of Operate switch.
 - a. No "OSC" indication on meter. Exciter-auxiliary plug P-3 is not installed, or jumpered incorrectly.
 - b. No "BUFFER" current indication. Drive control either fully counter-clockwise or open.
 - c. No "grid" current indication. Check position of crystal - VFO switch. Check crystals, check setting of drive control. Check the emission of the buffer by switching to "BUFFER" position and advancing drive control fully clockwise. Current should go to 40 ma. Check clamper tube as shorted clamper tube will short out final grid.
 7. Final Tuning. "Phone" position of Operate switch.
 - a. Unable to load final. Check ground leads, antenna leads, antenna change-over relay leads, and proper functioning of the relay. Check action of bandswitch rear wafer SW3D.
 - b. Read and understand the section "D Pi Network Tuning and Harmonic Suppression".
 - c. Check the final plate and screen voltage. Defective clamper tube will keep final from operating. Also check drive to final - normal grid drive is 2 ma.
 8. Reports of excessive harmonics of spurious signals; Phone or CW.
 - a. Read discussion on providing a good ground and pi-network tuning and harmonic suppression.
 - b. Overmodulating transmitter on phone can cause spurious signals.
 - c. Poor crystals.
 9. Reports of signals 20 to 60 kc on either side of carrier - Crystal operation. These spurious signals originate in the crystal. Some crystals will show some excitation near the fundamental mode of oscillation. Best solution to this problem is to replace the crystal.
 10. RF on chassis or microphone. Poor ground system or a very low impedance termination at the antenna connection at the transmitter may cause chassis and microphone to be hot. Read

- G. 10. the discussion on providing a good ground. Check for high standing wave ratio on the antenna feed line. Check that antenna is favorable to band used instead of to harmonically related band.
11. High "PLATE" current indication on meter CW-operation with key open. Check 6AQ5 clamper tube, and associated wiring.
12. Squeal or high modulator current indication on "PHONE" position.
- a. Check for acoustical feed back from receiver speaker or headphones.
 - b. Microphonic tube in the audio system.
 - c. Poor or intermittent ground connection to microphone or at the cable connector.
13. Adjustments.

a. Static Modulator Current

Turn the meter to "MOD". The reading should be between 55 and 70 ma. If it is out of these limits, turn the OPERATE switch to "OFF", pull out the AC plug, discharge the rear terminal of the high voltage bleeder R35 to the chassis with an insulated screwdriver and adjust the tap of R35 carefully (loosening the tap screw adequately to prevent breaking the resistance wire) toward the rear to increase the modulator current or toward the front to decrease it. Turn OPERATE switch to "TUNE" for warmup and again to "PHONE" to check the mod current.

b. Grid Drive Adjustment.

The grid drive on 10 and 6 meters may be adjusted by peaking L5 in the vicinity of 28 mc.

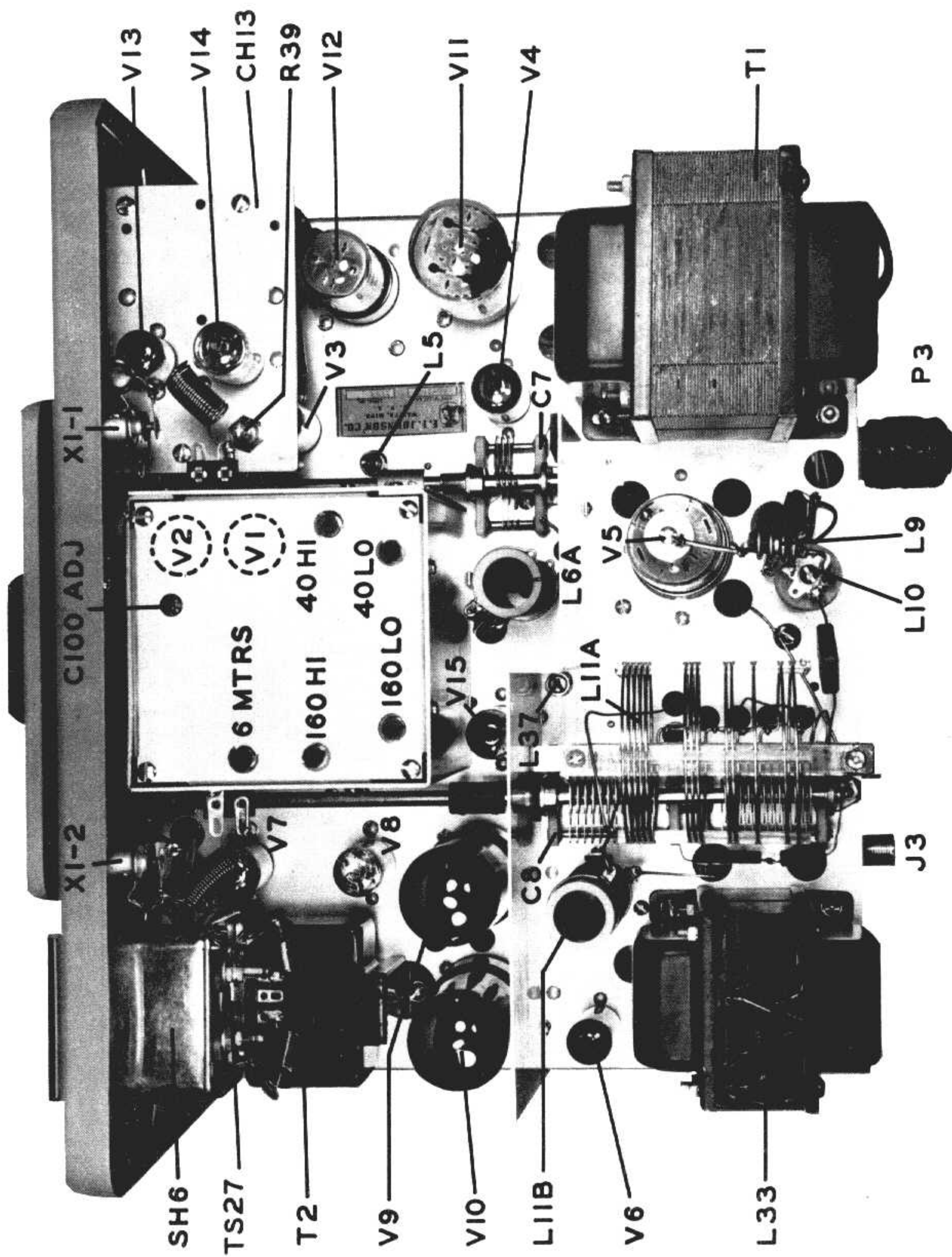


FIGURE I - COMPLETED CHASSIS TOP VIEW

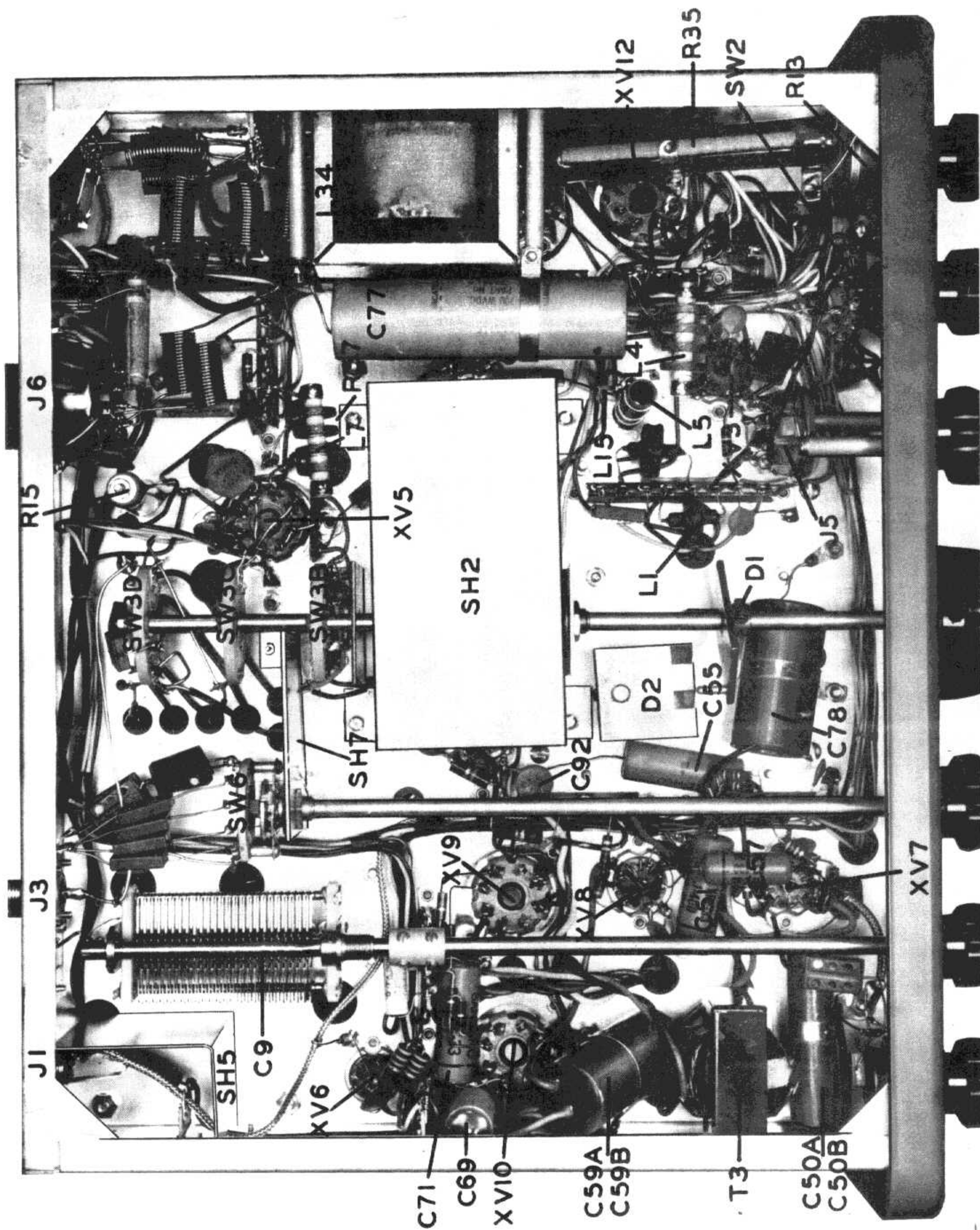


FIG. 2
COMPLETED CHASSIS BOTTOM VIEW

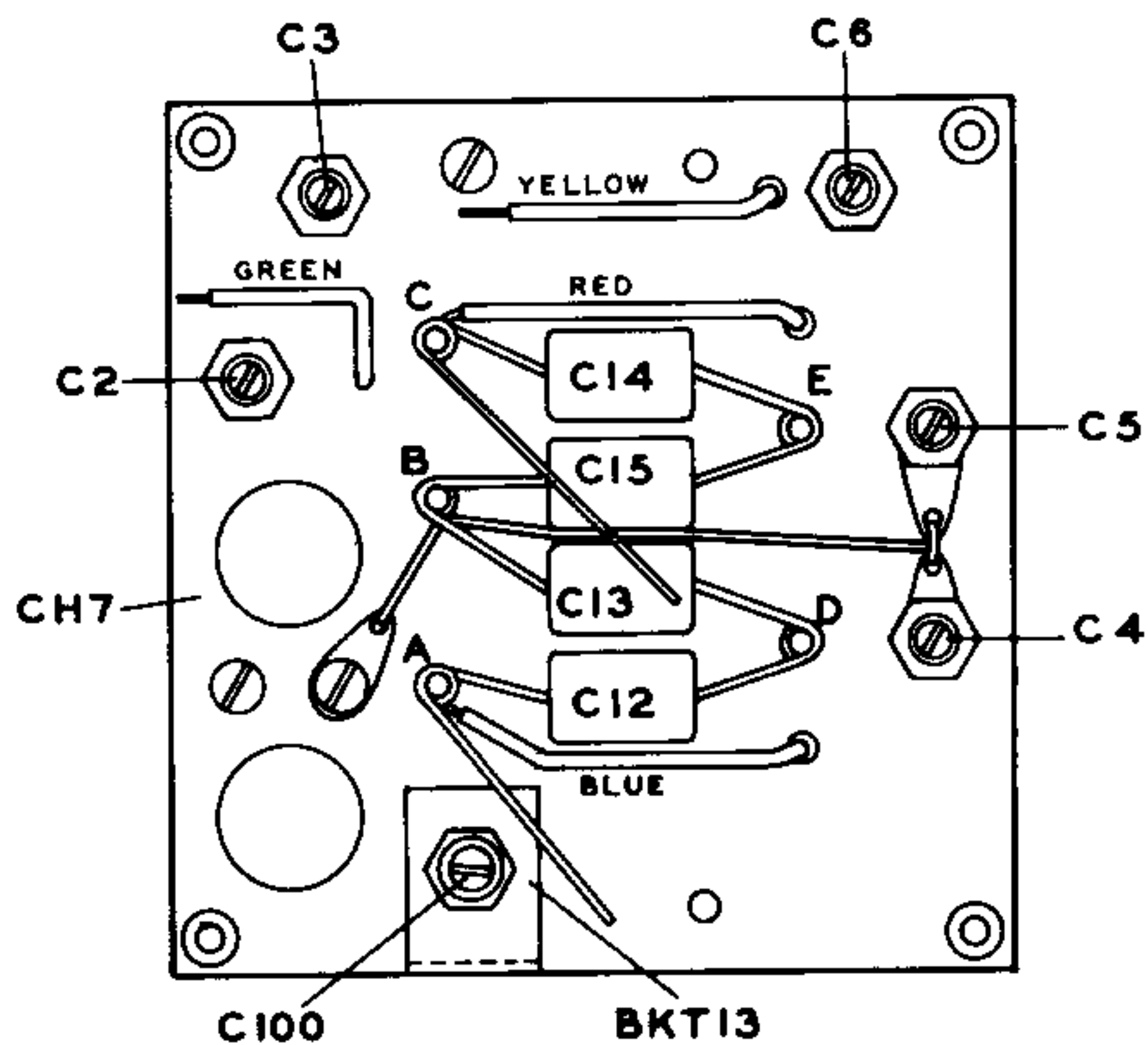


FIGURE 3A VFO-TOP VIEW

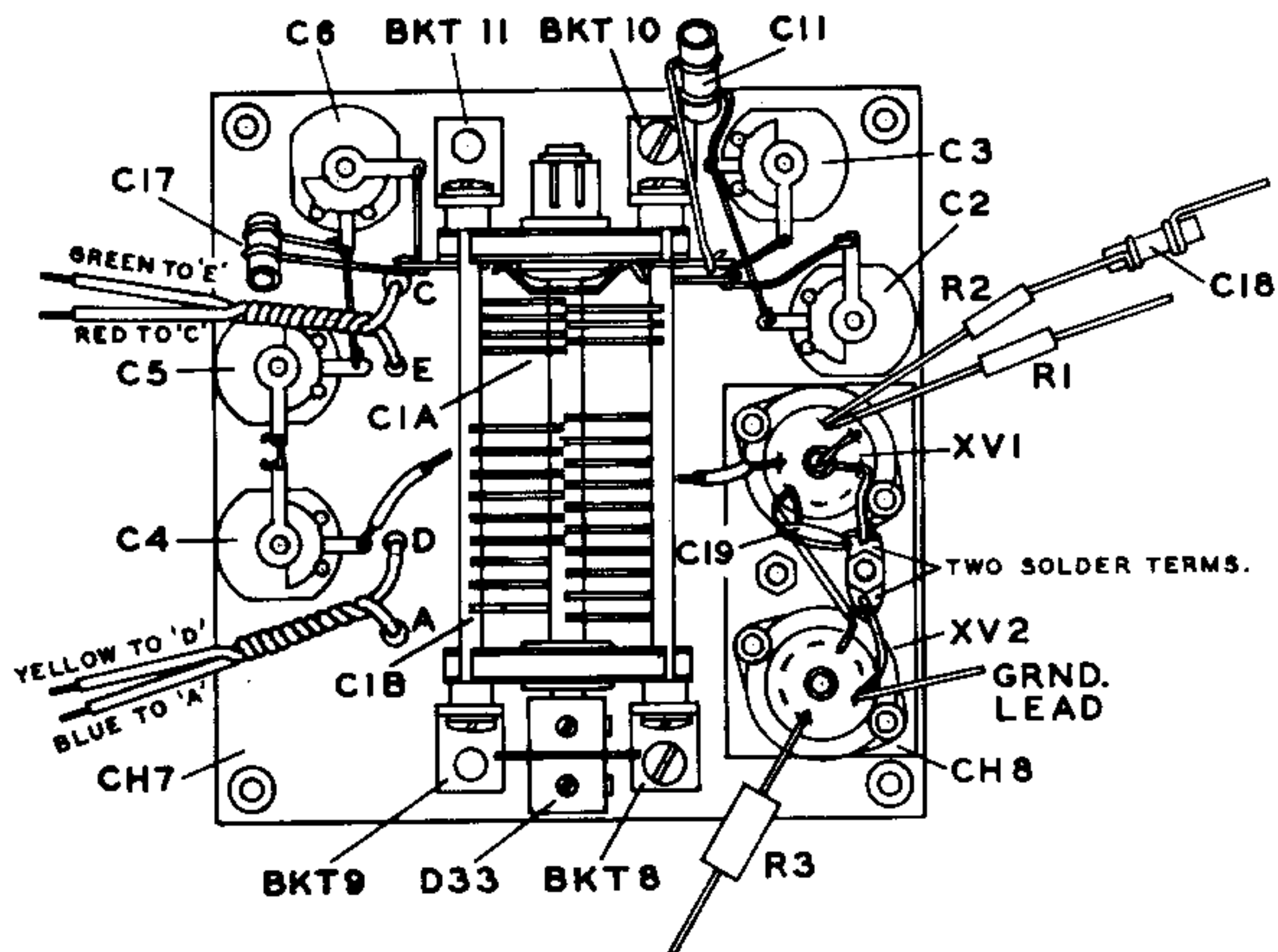


FIGURE 3B VFO-BOTTOM VIEW

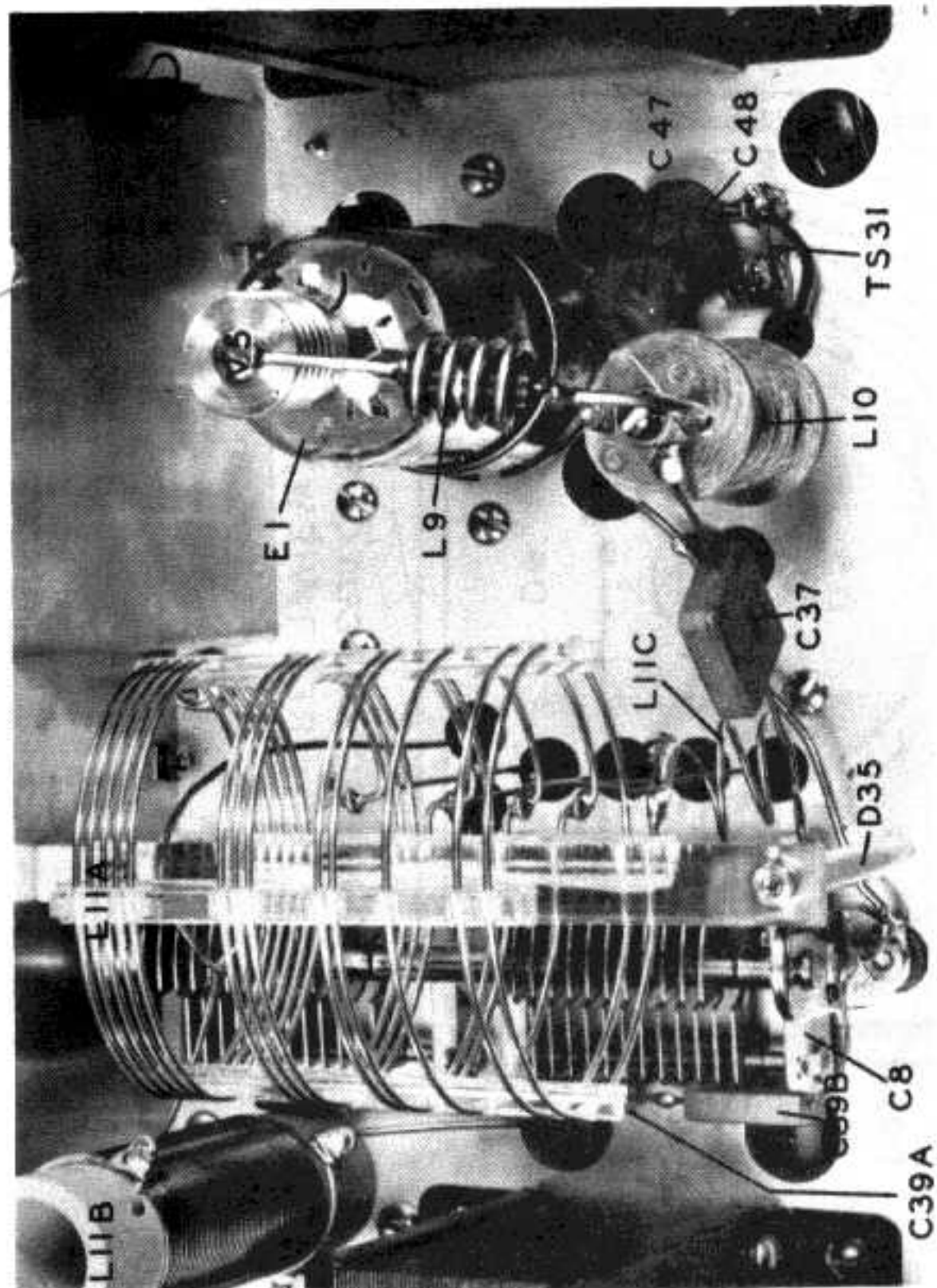


FIGURE 5B RF FINAL TOP VIEW

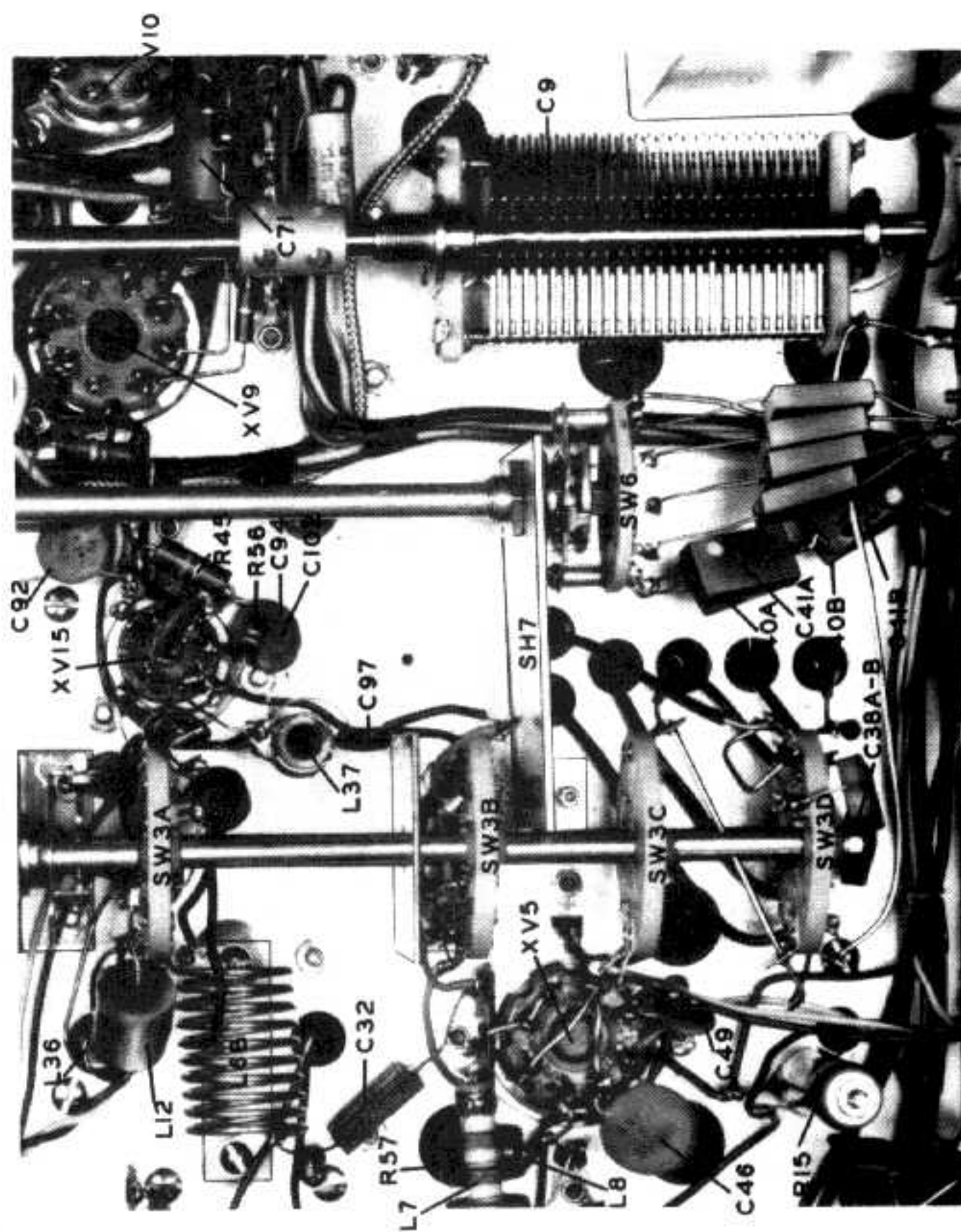


FIG 5C-RF FINAL BOTTOM VIEW

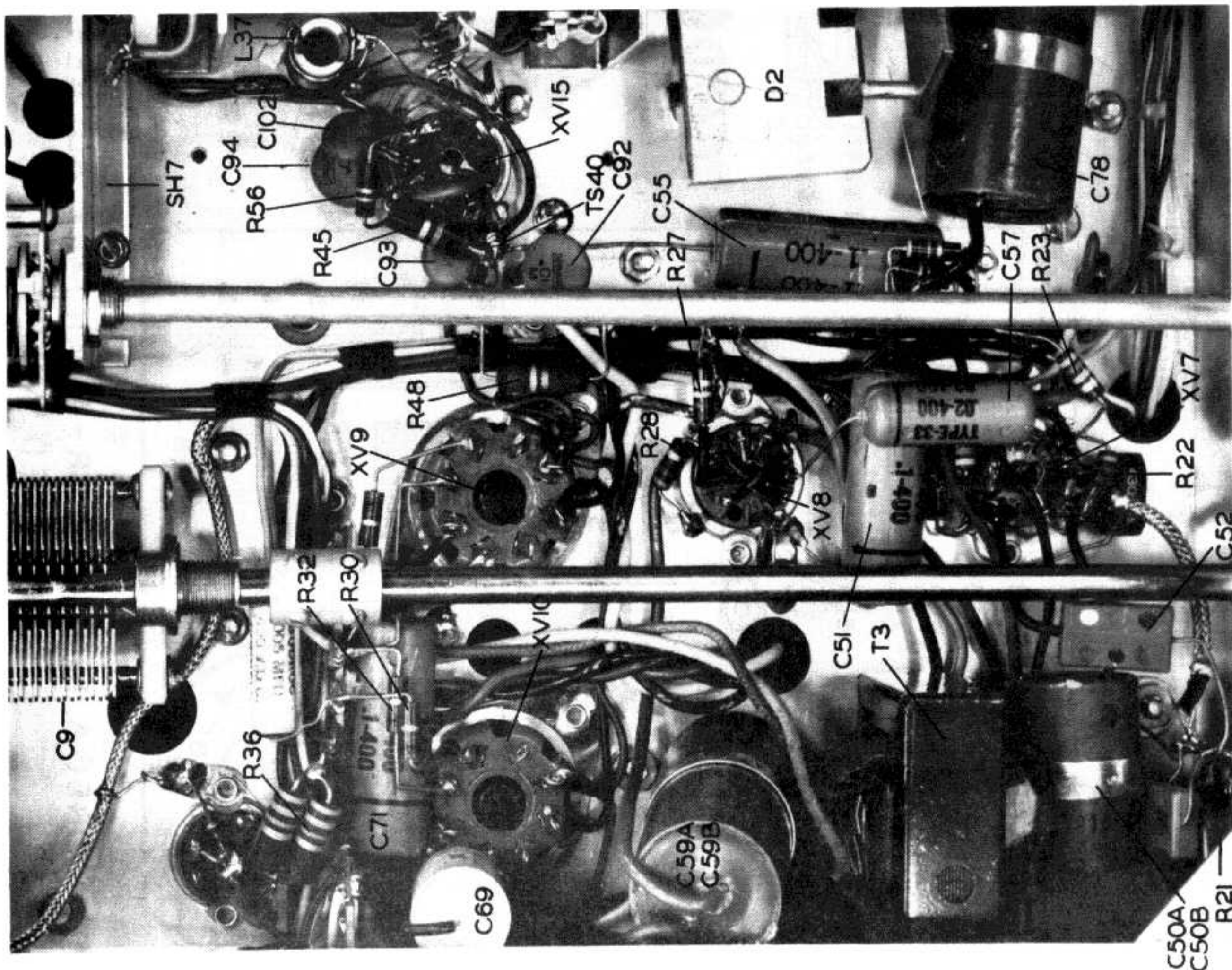
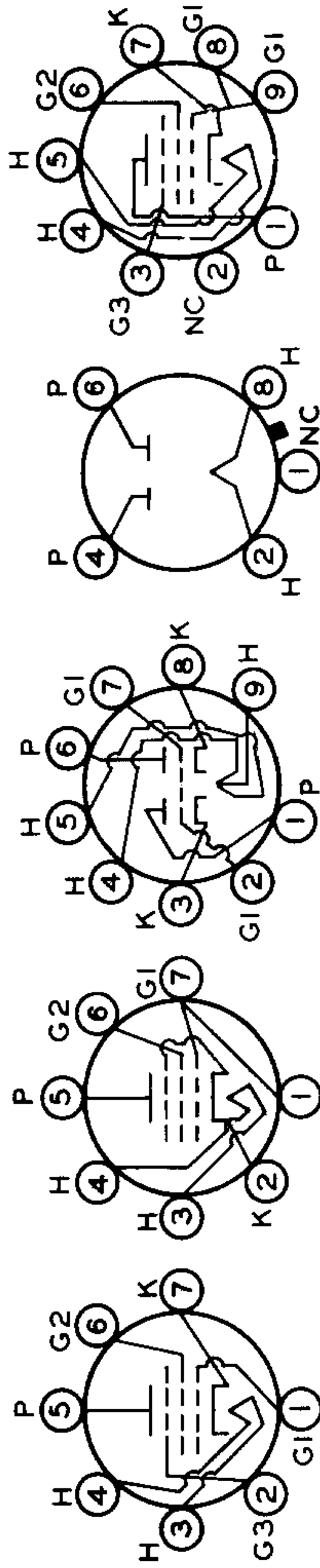
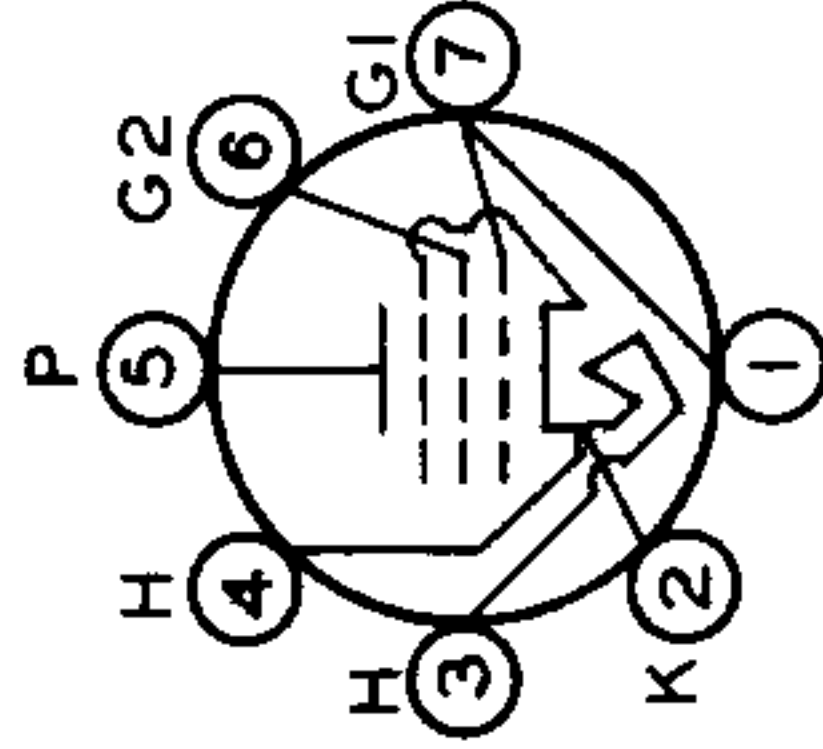


FIG. 5A
AUDIO SECTION

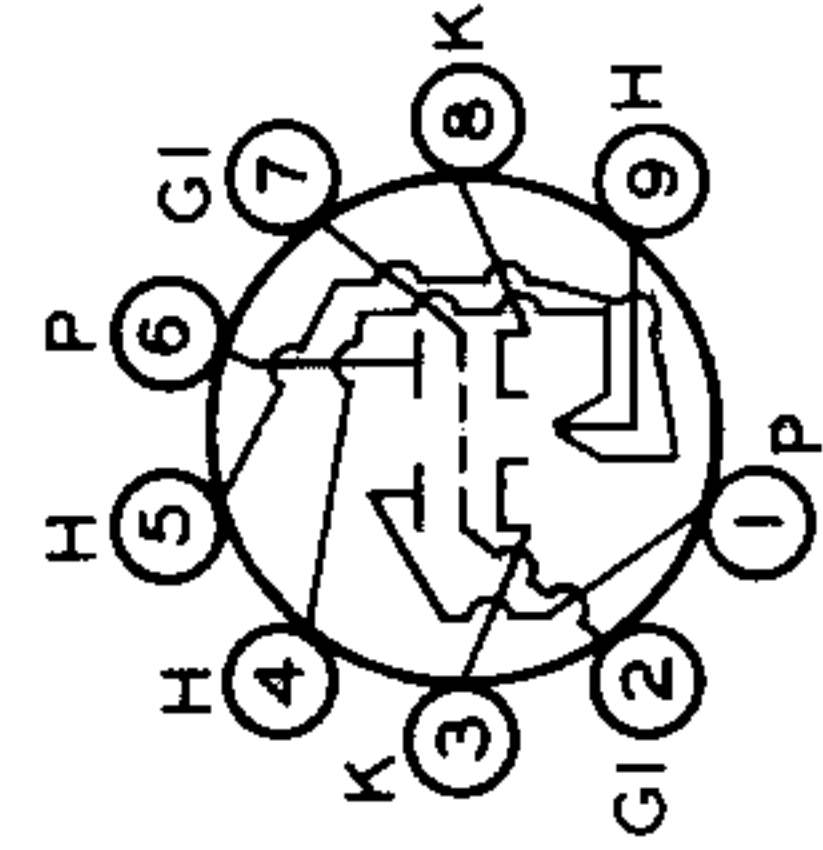
TUBE SOCKET CONNECTIONS-BOTTOM VIEW



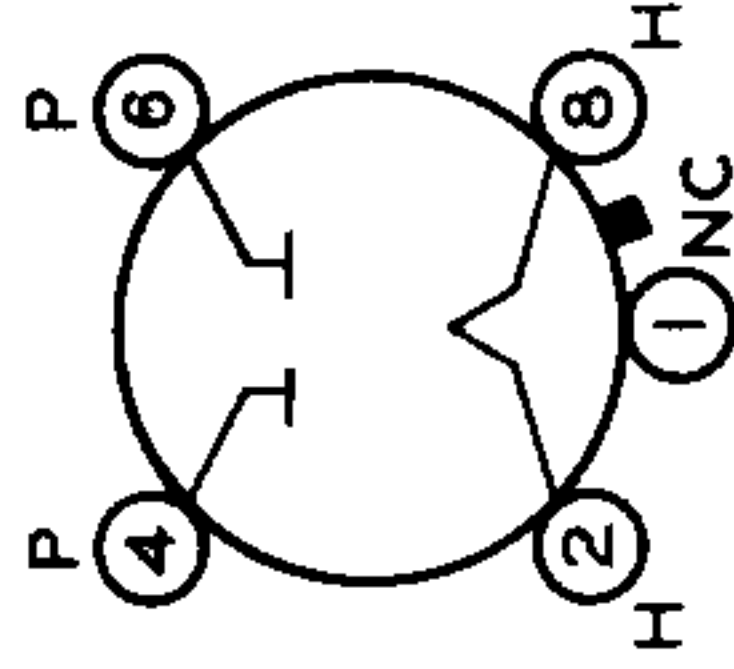
6AU6



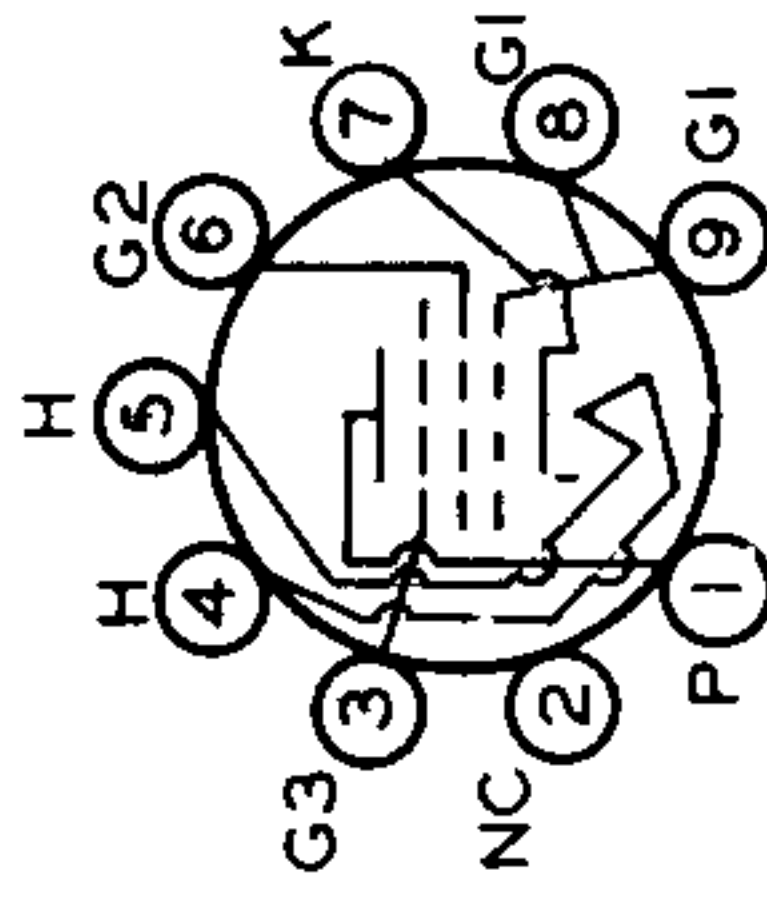
6AQ5



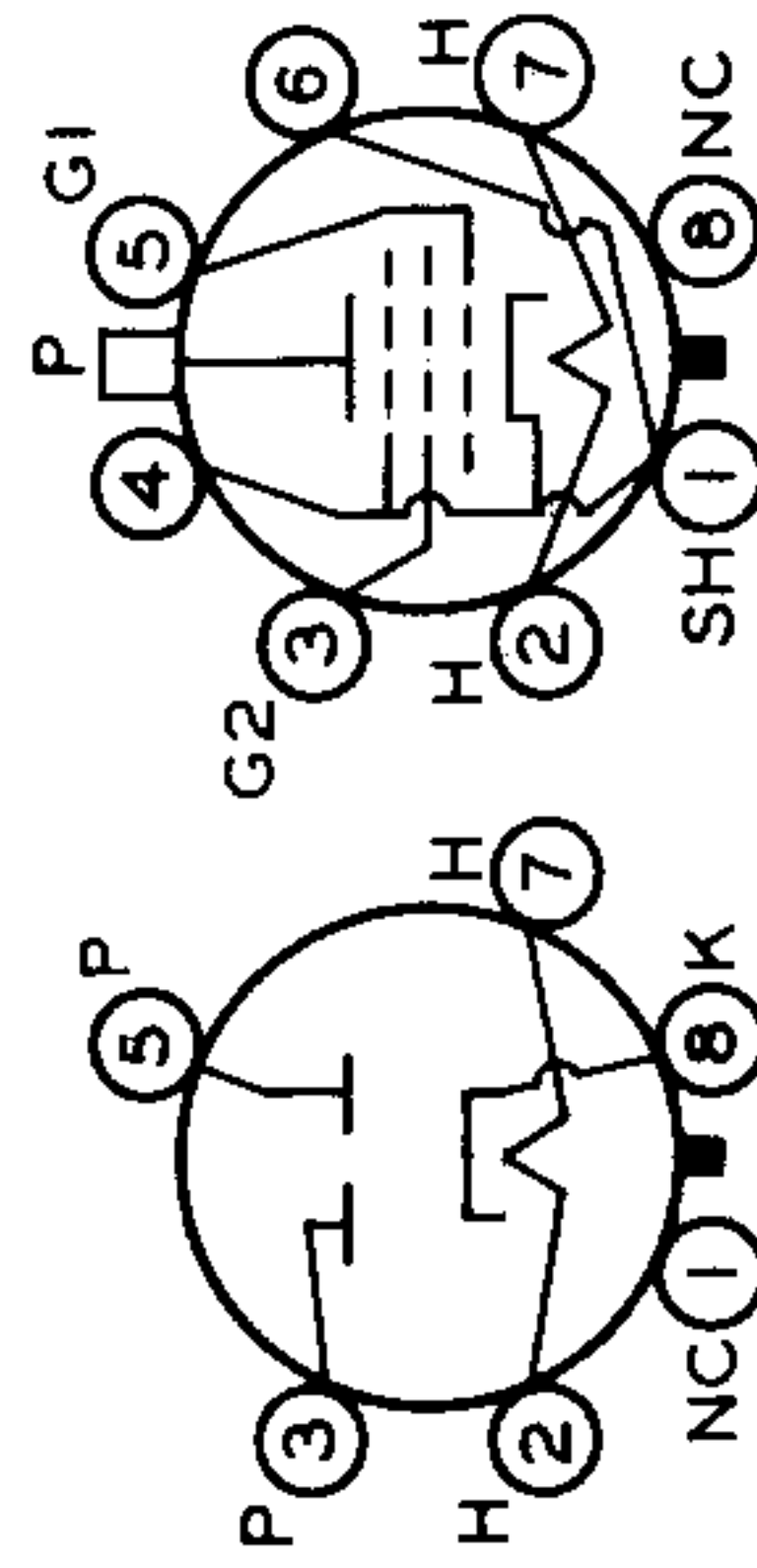
12AU7-12AX7



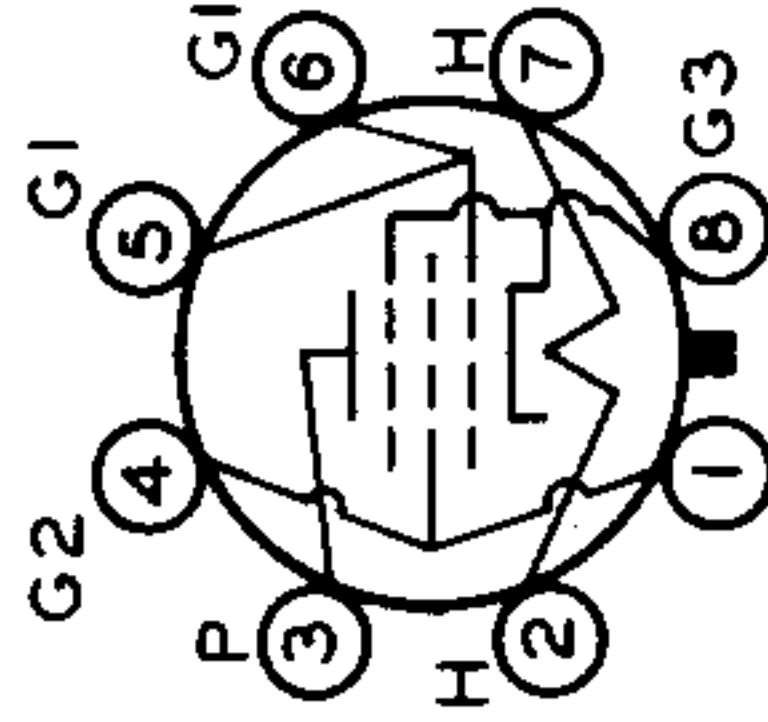
5R4GY



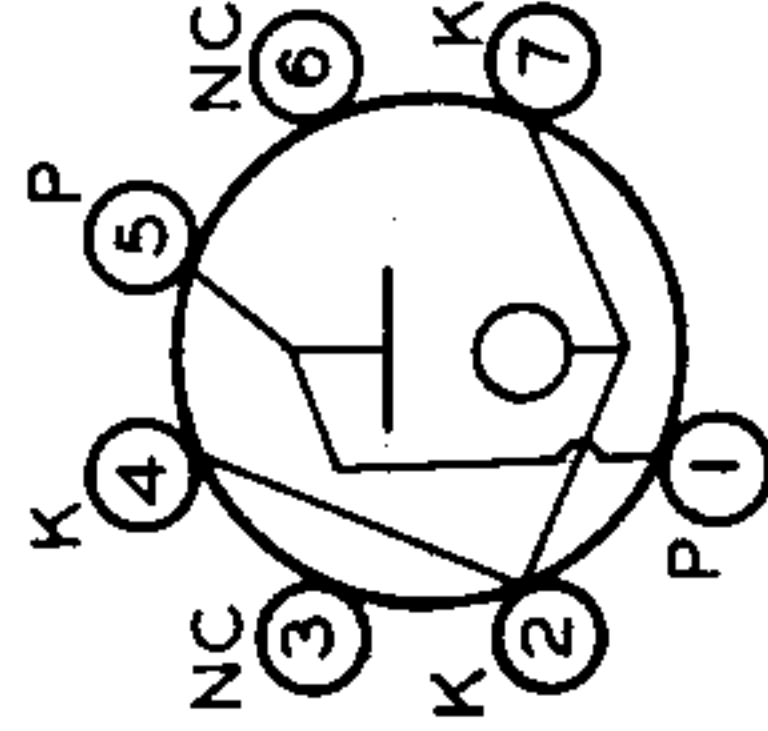
5763



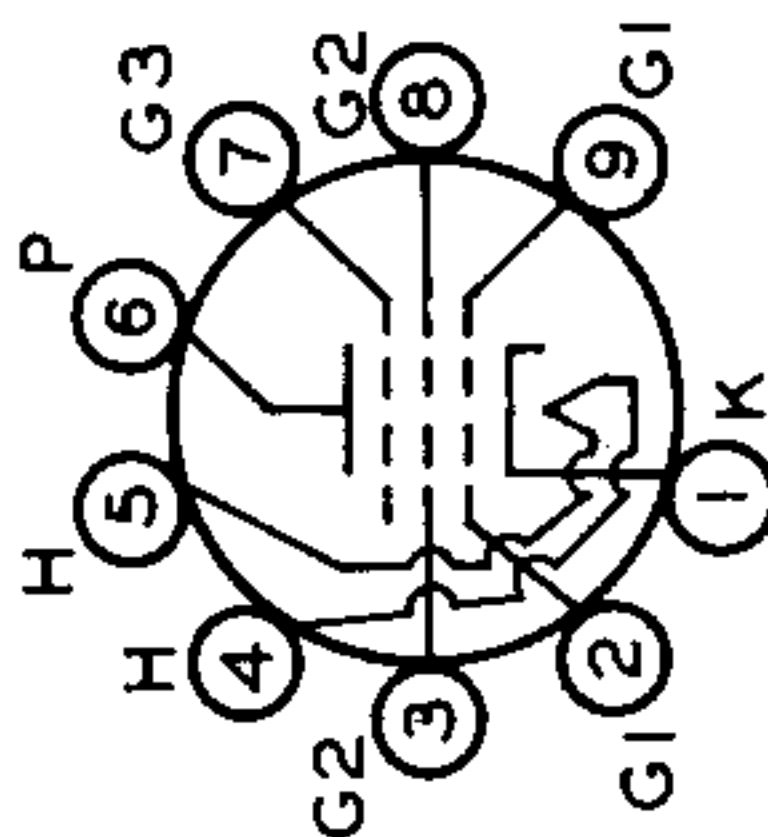
6AX5



7027



6CL6



6AX5

P-PLATE K-CATHODE H-HEATER SH-SHIELD OR SHELL
 GI-CONTROL GRID G2-SCREEN GRID
 G3-SUPPRESSOR GRID NC-NO INTERNAL CONNECTION

FIGURE 7A

COLOR	SIGNIFICANT FIGURE	DECIMAL MULTIPLIER	TOLERANCE %	VOLTAGE RATING *
BLACK	0	1	-	-
BROWN	1	10	1	100
RED	2	100	2	200
ORANGE	3	1,000	3	300
YELLOW	4	10,000	4	400
GREEN	5	100,000	5	500
BLUE	6	1,000,000	6	600
VIOLET	7	10,000,000	7	700
GRAY	8	100 MEG. (.01%)	8	800
WHITE	9	1000 MEG. (.1%)	9	900
GOLD	-	0.1	5	1000
SILVER	-	0.01	10	2000
NO COLOR	-	-	20	500

* APPLIES TO CONDENSERS ONLY.

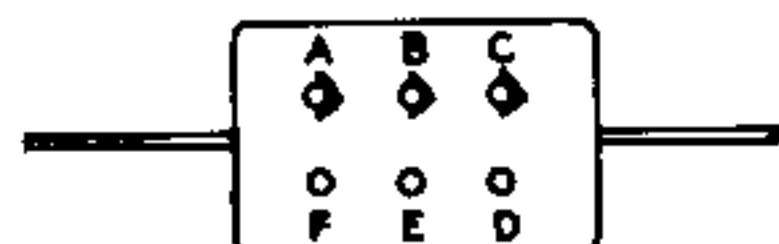


COLOR CODING OF FIXED RESISTORS

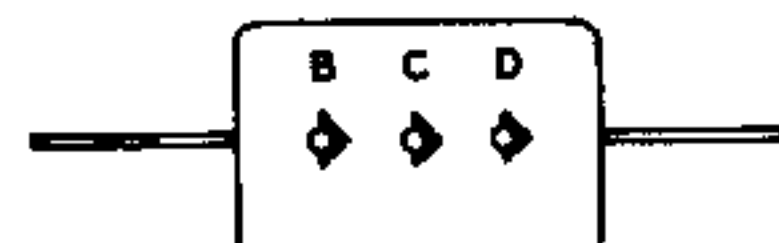
- A - FIRST SIGNIFICANT FIGURE OF RESISTANCE IN OHMS.
- B - SECOND SIGNIFICANT FIGURE.
- C - DECIMAL MULTIPLIER.
- D - RESISTANCE TOLERANCE IN PERCENT. IF NO COLOR SHOWN TOLERANCE IS $\pm 20\%$.

COLOR CODING OF FIXED CONDENSERS

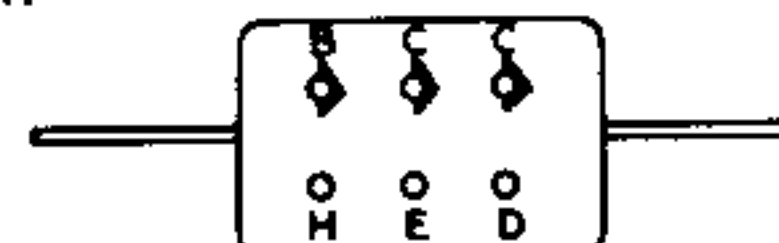
- A - TYPE: MICA BLACK, PAPER SILVER.
- B - FIRST SIGNIFICANT FIGURE OF CAPACITY.
- C - SECOND SIGNIFICANT FIGURE.
- D - DECIMAL MULTIPLIER.
- E - TOLERANCE.
- F - CHARACTERISTIC.
- G - THIRD SIGNIFICANT FIGURE.
- H - VOLTAGE RATING.



JAN FIXED CAPACITORS



RMA 3-DOT CODE 500VOLT $\pm 20\%$



RMA 6-DOT CODE



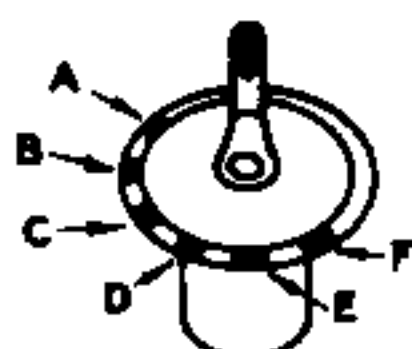
5-DOT RADIAL LEAD CERAMIC CAPACITOR

- A - TEMP. COEFFICIENT.
- B - FIRST SIGNIFICANT FIGURE.
- C - SECOND SIGNIFICANT FIGURE.
- D - DECIMAL MULTIPLIER.
- E - TOLERANCE.



MOLDED PAPER TUBULAR CAPACITOR

- A - FIRST SIGNIFICANT FIGURE.
- B - SECOND SIGNIFICANT FIGURE.
- C - DECIMAL MULTIPLIER.
- D - TOLERANCE.
- E - FIRST SIGNIFICANT VOLTAGE FIGURE.
- F - SECOND SIGNIFICANT VOLTAGE FIGURE.



BUTTON SILVER MICA CAPACITOR

- A - CLASS
- B - TOLERANCE
- C - MULTIPLIER
- D - THIRD DIGIT
- E - SECOND DIGIT
- F - FIRST DIGIT



4-DOT DISC CERAMIC RMA CODE



DIPPED MICA CAPACITOR

THE TOLERANCE RATING OF CAPACITORS IS DETERMINED BY THE COLOR CODE. FOR EXAMPLE: RED = 2%, GREEN = 5%, ETC. THE VOLTAGE RATING OF CAPACITORS IS OBTAINED BY MULTIPLYING THE COLOR VALUE BY 100. FOR EXAMPLE: ORANGE = 3X100 OR 300 VOLTS. BLUE = 6X100 OR 600 VOLTS.

THE PHYSICAL SIZE OF CARBON RESISTORS IS DETERMINED BY THEIR WATTAGE RATING. CARBON RESISTORS MOST COMMONLY USED ARE 1/2 WATT. HIGHER WATTAGE RATED RESISTORS WHEN SPECIFIED ARE PROGRESSIVELY LARGER PHYSICALLY.

STANDARD COLOR CODE
FIGURE 7B

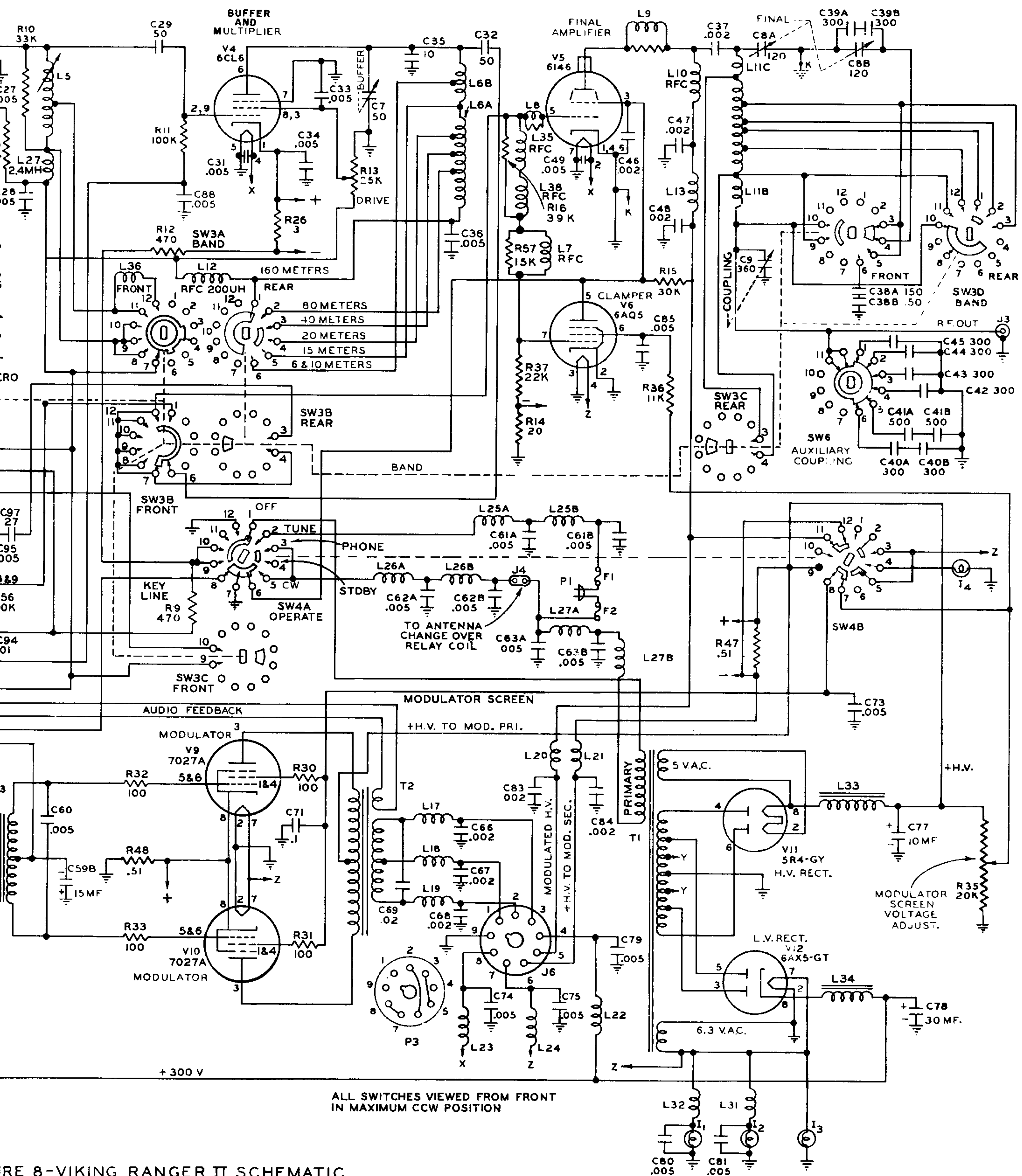


FIGURE 8-VIKING RANGER II SCHEMATIC

Viking Ranger II Transmitter

Parts List

Part No. or Drawing No.	Item No.	Qty.	Description
22.1155-2	BKT1	1	Meter shield bracket
16.1001-1	BKT2	1	Underchassis coupling switch bracket
16.1165-1	BKT3	1	Underchassis band switch bracket
16.1001-4	BKT4-7	4	Topchassis component bracket
16.1167	BKT8-12	5	VFO capacitor bracket
17.1171	BKT13	1	Differential capacitor bracket
169-26	C1A, B	1	Special LA type dual variable capacitor
160-107-51	C2, 5	2	15M11 variable capacitor
160-130-50	C3, 6	2	30M8 variable capacitor
160-107-50	C4	1	15M11 variable capacitor
149-3-3	C7	1	50R12 variable capacitor
149-530-3	C8A, F	1	120RD18 variable capacitor
149-13-3	C9	1	360R12 variable capacitor
22.1740	C10A	1	15 mmf. N330 disc ceramic capacitor
22.954	C11	1	62 mmf. 2 1/2% NPO 500 W.V. tube ceramic cap.
22.804	C12, 13	2	500 mmf. 2% mica capacitor
22.805	C14, 15	2	1,000 mmf. 2% mica capacitor
22.809	C16	1	91 mmf. 2 1/2% N080 500 W.V. tube ceramic cap.
22.823	C17	1	140 mmf. 2 1/2% N080 500 W.V. tube ceramic cap.
22.807	C18	1	43 mmf. 2 1/2% N080 500 W.V. tube ceramic cap.
22.827	C19-21, 25-28, 30, 31, 33, 34, 36, 49, 61A, 61B, 62A, 62B, 63A, 63B, 70, 72-75, 79, 80, 81, 82, 85, 87, 88, 95, 96, 102	36	.005 at 600 W.V. ceramic disc capacitor
22.774	C22, 40A, 40B, 42, 43, 44, 45	7	300 mmf. 20% 500 W.V. molded mica capacitor
22.1739	C10B	1	18 mmf. NPO disc ceramic capacitor
22.4061-10	C58	1	330 mmf. 10% dipped mica capacitor
22.4047-5	C101	1	82 mmf. 5% dipped mica capacitor
22.777	C23	1	25 mmf. 5% at 500 W.V. silver mica capacitor
22.862	C24, 53	2	200 mmf. 20% at 500 W.V. molded mica capacitor
22.773	C29, 32	2	50 mmf. 20% at 500 W.V. molded mica capacitor
22.856	C35	1	10 mmf. 5% at 500 W.V. silver mica capacitor
22.956	C46-48, 64-68, 83, 84	10	.002 GMV 1500 W.V. ceramic disc capacitor
22.955	C37	1	.002 mfd 20% 1500 W.V. molded mica capacitor
22.957	C38A, B	2	150 mmf. 5% 500 W.V. silver mica capacitor
22.859	C39A, B	2	300 mmf. 20% 500 W.V. molded mica capacitor
22.958	C41A, B, 52, 86	4	500 mmf. 20% 500 W.V. molded mica capacitor
22.4065-10	C56	1	470 mmf. 10% dipped mica capacitor
22.768	C51, 55, 71	3	.1 mfd. +100 -20% 400 W.V. paper tube capacitor
22.767	C57	1	.02 mfd. 20% 400 W.V. paper tube capacitor
22.765	C59A, B, 90A, B	2	15-15 mfd. +100 -20% 150 W.V. electrolytic cap.
22.960	C50A, B	1	10-10 mfd. +100 -20% 50 W.V. electrolytic cap.
22.963	C60	1	.005 mfd. 20% 400 W.V. paper tube capacitor
22.964	C69	1	.02 mfd. 20% 1600 W.V. paper tube capacitor
22.961	C77	1	10 mfd. +100 -20% 700 W.V. electrolytic capacitor
22.962	C78	1	30 mfd. +100 -20% 450 W.V. electrolytic capacitor

Viking Ranger II Transmitter

Parts List

Part No. or Drawing No.	Item No.	Qty.	Description
22.1111	C89	1	.05 mfd. 20% 200 W.V. paper tube capacitor
22.4033-5	C91	1	22 mmf. 5% 500 W.V. dipped mica capacitor
22.1097	C93, 94, 54, 92	4	.01 mfd. $\pm 100 \pm 20\%$ 600 W.V. ceramic disc cap.
22.4035-10	C97	1	27 mmf. 10% 500 W.V. dipped mica capacitor
22.1494	C98	1	12 mmf. NPO 10% 600 W.V. disc ceramic cap.
22.1495	C99	1	12 mmf. N750 1% 600 W.V. disc ceramic cap.
160-311-50	C100	1	19 mmf. Max. variable capacitor
17.848	CH1	1	Chassis
23.1055-3	CH2	1	Cabinet
23.1056-4	CH3	1	Panel
17.853	CH4	2	Chassis Rails
17.820	CH5	1	VFO top
17.819	CH6	1	VFO side plate
18.699	CH7	1	VFO phenolic plate
17.855	CH8	1	VFO subchassis
22.1475-2	CH11	4	Button, Polyethylene Rest
22.1001	CH12	1	3/16" round Metal-tex gasket
17.931	CH13	1	Keyer plate
23.1059	D1	1	Drive arm for VFO switch
14.504	D2	1	Drive cam for VFO switch
23.1062	D3	1	Planetary drive assembly
17.858-5	D4	1	Dial escutcheon
22.993-4	D5	1	Dial plate
042.0361-310	D6	24 1/4"	Adhesive backed gasket strip 15/32" wide
22.995	D7	5 in.	Rubber light blocks
23.1064-003	D8	1	Dial pointer
23.564-74	D9A	1	Red jewel
23.907-22	D10, 11	2	100-0 dial
116.223-11	D12-18	7	Phenolic knobs
116.263-11	D19	1	1 5/8" phenolic knob
23.1060-2	D20	1	Crystal knob cover
23.1463	D21	1	2 3/8" phenolic knob
13.123-12	D22	5	3/8" -32 panel bearing
13.760-2	D23	2	Coupling
13.49-9	D24	2	VFO subchassis spacer
14.31-62	D25	3	1 3/8" crystal socket spacer
13.49-28	D26	2	Capacitor spacer
14.31-64	D27	4	2 1/8" VFO chassis rods
14.31-65	D28, 36, 37, 38	7	2 15/16" VFO chassis rods
14.139-3	D29	1	7 1/4" extension shaft
14.139-2	D30	1	6 7/16" extension shaft
14.139-1	D31	1	5 5/16" extension shaft
18.638-2	D32	5	VFO trimmer shafts
104-264-3	D33	1	Insulated shaft coupler
23.564-75	D9B	1	Green jewel
18.666-4	D34	1	Insulated coupling
13.51-4	D35	3	Final and HV capacitor spacer
22.1210-1	E4	1	Iron core

Viking Ranger II Transmitter

Parts List

Part No. or Drawing No.	Item No.	Qty.	Description
13.904-3	E1	1	Tube cap
22.950	E2	2	Pinout 513 coil fastener
18.36-7	E5	2	Fiber shoulder washers
22.982	F1	1	3.2 amp. 125V MDL fuse
22.742	F2	1	5 amp. 250V MTH fuse
22.1171	E3	1	Coil fastener (3/8" dia., #4-40 tap)
22.113-1	G1-6	6	9.16" O.D. grommet
22.113-5	G7	1	11.32" O.D. grommet
26.236	H1	1	Right section harness
26.237	H2	1	Left section harness
			#4 hardware envelope
			#6 hardware envelope
			#8 hardware envelope
			#10 hardware envelope
22.377	J1-4	4	#51 pilot lamp
22.979	J1	1	Microphone jack
22.980	J2	1	Key jack
22.746	J3	1	83-1R coaxial receptacle
126-105	J4	1	Relay jack
22.849-2	J5	1	Crystal (8 pin octal) socket
2.997	J6	1	Power socket (9 pin)
23.968-2	L1	1	Dual VFO coil
22.844-2	L2	1	VFO output coil 52 UH
22.951	L3, 4, 7	3	RFC Miller 4537
22.949	L5	1	Oscillator coil
23.902-5	L6A	1	L.F. buffer coil
23.913-2	L6B	1	H.F. buffer coil
23.1212-2	L8	1	Parasitic suppressor
23.1329-2	L9	1	Parasitic suppressor
102-752-5	L10	1	RFC 1 MH.
22.844-1	L12, 16	2	RFC 200 UH.
23.1061	L11A	1	Main final coil
23.902-4	L11B	1	160 meter final coil
16.1484	L11C	1	6 meter final coil
22.1193	L27	1	2.4 MH RFC.
23.1000	L13, 14, 17-22	8	4.7 UH RFC.
22.844-3	L15	1	100 UH RFC.
23.1175-28	L37	1	6 meter doubler coil ceramic
22.987	L33	1	10 Henry H.V. choke SNC 2428
22.749	L34	1	15 Henry L.V. choke SNC P1501
22.1549	L35, 38	2	20 UH RFC.
22.1768	M1		0-5MA D.C. Milliammeter
22.981	P1	1	Fused plug
23.1031	P2	1	Relay plug
22.978	P3	1	9 pin "Octal" mica filled plug

Viking Ranger II Transmitter

Parts List

Part No. or Drawing No.	Item No.	Qty.	Description
22. 5097-10	R1, 11, 42, 45, 46	6	100K ohm 1/2 watt carbon resistor
22. 5019-10	R2	1	56 ohm 1/2 watt carbon resistor
22. 7079-10	R3	1	18K ohm 2 watt carbon resistor
22. 6053-10	R4	1	1.5K ohm 1 watt carbon resistor
22. 5041-10	R5, 12, 9	3	470 ohm 1/2 watt carbon resistor
22. 5081-10	R6, 38	2	22K ohm 1/2 watt carbon resistor
22. 5093-10	R8	1	68K ohm 1/2 watt carbon resistor
22. 5085-10	R10	1	33K ohm 1/2 watt carbon resistor
22. 732	R13	1	25K ohm 4 watt W. W. potentiometer
22. 8056-5	R14	1	20 ohm 1/2 watt carbon resistor 5%
22. 972	R15	1	30K ohm 20 watt W. W. fixed resistor
22. 6063-10	R16	1	3.9K ohm 1 watt carbon resistor
22. 5065-10	R18, 22, 52, 53	4	4.7K ohm 1/2 watt carbon resistor
22. 5121-10	R17, 40	2	1 megohm 1/2 watt carbon resistor
22. 5113-10	R19, 23	2	470K ohm 1/2 watt carbon resistor
22. 5105-10	R20	1	220K ohm 1/2 watt carbon resistor
22. 832	R21	1	1 megohm 1/2 watt potentiometer 30%
22. 5089-10	R24, 43	2	47K ohm 1/2 watt carbon resistor
22. 8036-5	R25, 26	2	3 ohm 1/2 watt W. W. resistor 5%
22. 5101-10	R27	1	150K ohm 1/2 watt carbon resistor
22. 5047-10	R28	1	820 ohm 1/2 watt carbon resistor
22. 5025-10	R30-33	4	100 ohm 1/2 watt carbon resistor
22. 7069-10	R54	1	6.8K ohm 2 watt carbon resistor
22. 973	R35	1	20K ohm 50 watt Adj. W. W. resistor
22. 6085-10	R36A, B, C	3	33K ohm 1 watt carbon resistor
22. 6073-10	R45	1	10K ohm 1 watt carbon resistor
22. 6081-10	R37	1	22K ohm 1 watt carbon resistor
22. 1115	R39	1	100K 1/2 watt carbon resistor
22. 5079-10	R41	1	18K ohm 1/2 watt carbon resistor
22. 5109-10	R44	1	330K ohm 1/2 watt carbon resistor
22. 8018-5	R47, 48	2	.51 ohm 1/2 watt W. W. resistor 5%
22. 6077-10	R57	1	15000 ohm 1 watt resistor
17. 847	SH1	1	Final shield
17. 857	SH2	1	Buffer shield
17. 846	SH3	1	VFO shield
22. 948-2	SH4	1	Socket shield
17. 856	SH5	1	Mic. & key shield
22. 1154-11	SH6A	1	Meter shield
17. 1206	SH7	1	Loading switch mounting shield
22. 988	SW1	1	VFO bandswitch
22. 989	SW2	1	VFO-Crystal switch
22. 1667	SW3	1	Bandswitch
22. 991	SW4	1	Operate switch
22. 847-2	SW5	1	Meter switch
22. 992	SW6	1	Coupling switch
22. 985	T1	1	Power transformer SNC 2793
22. 986	T2	1	Modulation transformer SNC P2429
22. 753	T3	1	Driver transformer SNC 1503
22. 1401-1	TS30, 35	2	3 lug terminal strip 1 mounting foot
22. 740-3	TS24-27	4	3 lug terminal strip

Viking Ranger II Transmitter

Parts List

<u>Part No. or Drawing No.</u>	<u>Item No.</u>	<u>Qty.</u>	<u>Description</u>
22.740-4	TS15, 16, 19, 20	4	4 lug terminal strip
22.740-5	TS18, 19, 21, 23, 28, 29	6	5 lug terminal strip
22.740-6	TS17, 37	2	6 lug terminal strip
22.740-7	TS22	1	7 lug terminal strip
22.740-8	TS34	1	8 lug terminal strip
22.837	TS31, 32, 33, 39, 40	5	2 lug terminal strip
22.780	V1	1	6AU6 electron tube
22.787	V2	1	OA2 electron tube
22.1118	V3, 4	2	6CL6 electron tube
22.788	V5	1	6146 electron tube
22.781	V6	1	6AQ5 electron tube
22.915	V7	1	12AX7 electron tube
22.916	V8, 14	2	12AU7 electron tube
22.1668	V9, 10	2	7027 electron tube
22.784	V11	1	5R4GY electron tube
22.1123	V12	1	6AX5GT electron tube
22.786	V13	1	6AL5 electron tube
22.1248	V15	1	5763 electron tube
71.91-100	W1	15 1/4 ft.	#20 black plastic covered tinned copper wire
71.91-105	W2	9 ft.	#20 green plastic covered tinned copper wire
71.91-104	W3	5 3/4 ft.	#20 yellow plastic covered tinned copper wire
71.91-102	W4	4 1/4 ft.	#20 red plastic covered tinned copper wire
71.91-106	W5	2 ft.	#20 blue plastic covered tinned copper wire
71.27-115	W6	6 ft.	#16 bare tinned copper wire
71.49-114	W7	83"	Black line cord 18-2 POSJ 1/64 type
71.13-125	W8	36 ft.	#18 Formex copper wire
71.91-103	W9	1 1/2 ft.	#20 orange plastic covered tinned copper wire
42.24-050	W12	1 ft.	.053 ID varnished tubing
42.24-107	W13	1/2 ft.	.133 ID tubing
22.997	W15	1/2 ft.	3/8" round wood dowling
42.24-113	W16	1"	.208 ID black Vinylite tubing
71.91-108	W17	1 1/2 ft.	#20 gray plastic covered tinned copper wire
71.91-109	W18	1 1/2 ft.	#20 white plastic covered tinned copper wire
23.1047	XI1, 2	2	Pilot light assembly
23.566-3	XI3, 4	2	Socket assembly
22.975	XV1, 2, 6, 13	4	7 pin mica filled min. socket
22.976	XV3, 4, 7, 8, 14, 15	6	9 pin mica filled min. socket
22.849-2	XV5, 9-12	5	8 pin octal mica filled socket

